



OPA11HT

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2

OPERATIONAL AMPLIFIERS

Wide Temperature-Range General Purpose OPERATIONAL AMPLIFIER

FEATURES

- **-55°C TO +175°C SPECIFICATIONS**
- **30nA MAX. INPUT BIAS CURRENT AT +175°C**
- **±6mV, MAX. INPUT OFFSET VOLTAGE AT +175°C**
- **±5 μ V/°C TYP. INPUT OFFSET VOLTAGE COEFFICIENT**
- **12MHz BANDWIDTH, TYPICAL**
- **HERMETIC PACKAGE WITH STANDARD PINOUT (741-TYPE)**

DESCRIPTION

These specifications give you a versatile operational amplifier that will work in circuits that are subjected to extremely wide temperature ranges. Typical applications for OPA11HT include general purpose gain blocks, high-speed pulse amplifiers, audio amplifiers, high-frequency active filters, high-speed integrators, and photodiode amplifiers.

You're assured of this product's performance over the -55°C to +175°C range because we conduct 100% screening procedures in accordance with MIL-STD-883, method 5004, class B. Burn-in is performed at 200°C. Our sample and inspection procedures include both destructive and nondestructive bonding wire

pull tests in accordance with Method 2011 of MIL-STD-883. The product is assembled in a clean-room environment.

Model OPA11HT is internally compensated for stability at all gains. Pins are available for special tailoring of the bandwidth compensation. Significant advantages in high gain, wide bandwidth, low-bias current, high output current and high common-mode rejection are provided by OPA11HT. Inputs are protected against common-mode voltages up to the value of the power supplies while the output is current limited to offer short circuited protection. TO-99 hermetic package has standard 741-type pinout arrangement.

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PDS-476

SPECIFICATIONS

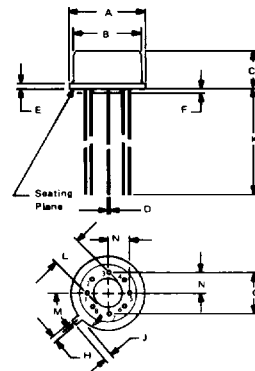
ELECTRICAL

Specifications at $\pm 15\text{VDC}$ and $T_A = +175^\circ\text{C}$ unless otherwise noted.

MODEL	OPA11HT				
CHARACTERISTIC	SYMBOL	MIN	TYP	MAX	UNIT
OPEN LOOP GAIN, DC, single-ended					
A_v					
No load		94	103		dB
$R_L = 2\text{k}\Omega$			100		dB
RATED OUTPUT					
Voltage, $R_L = 2\text{k}\Omega$	V_{om}	± 10	± 12		V
Current ($T_A = 25^\circ\text{C}$)	I_{om}	± 15	± 23		mA
DYNAMIC RESPONSE ($T_A = 25^\circ\text{C}$)					
Small-Signal Bandwidth (0dB)			12		MHz
Full-Power Bandwidth $V_{out} = \pm 10\text{V}$	BW_{fp}	50	75		kHz
Slew Rate $R_L = 2\text{k}\Omega$	SR	4	7		V/ μsec
Settling Time (0.1%)			1.5		μsec
Rise Time (10% to 90%, small-signal)			30		nsec
INPUT OFFSET VOLTAGE					
V_{io}					
Initial (without adj., at 25°C)			± 1	± 5	mV
Over Temperature					
$T_A = +175^\circ\text{C}$				± 6	mV
$T_A = -55^\circ\text{C}$				± 7	mV
Average V_{io} coefficient			± 5		$\mu\text{V}/^\circ\text{C}$
Average V_{io} coefficient vs supply voltage ($T_A = 25^\circ\text{C}$)			± 10	± 200	$\mu\text{V}/\text{V}$
INPUT BIAS CURRENT					
I_{ib}					
Initial at $+25^\circ\text{C}$			± 10	± 25	nA
Over Temperature					
$T_A = +175^\circ\text{C}$				± 30	nA
$T_A = -55^\circ\text{C}$				± 40	nA
Average I_{ib} coefficient			± 0.1		nA/ $^\circ\text{C}$
INPUT DIFFERENCE CURRENT					
I_{io}					
Initial at $+25^\circ\text{C}$			± 10	± 25	nA
Over Temperature					
$T_A = +175^\circ\text{C}$				± 30	nA
$T_A = -55^\circ\text{C}$				± 40	nA
Average I_{io} coefficient			± 0.1		nA/ $^\circ\text{C}$
INPUT IMPEDANCE ($T_A = 25^\circ\text{C}$)					
Differential	r_i C_i	100	300 3		M Ω pF
Common Mode	$r_i(\text{CM})$ $C_i(\text{CM})$		1000 3		M Ω pF
INPUT VOLTAGE RANGE					
Common Mode				± 11	V
Differential Mode				± 12	V
Common-Mode Rejection	CMR	80	100		dB
Over Temperature ($-55^\circ\text{C} \leq T_A \leq +175^\circ\text{C}$)			100		dB
POWER SUPPLY ($T_A = 25^\circ\text{C}$)					
Rated Voltage	V_{CC}			± 15	V
Voltage Range, derated			± 8 to ± 22		V
Current, quiescent	I_q		± 3	± 3.7	mA
Over Temperature ($-55^\circ\text{C} \leq T_A \leq +175^\circ\text{C}$)			± 3		mA
Power Supply Rejection					
Ratio ($T_A = +175^\circ\text{C}$)	PS_{rr}	80	100		dB
TEMPERATURE RANGE					
Specification		$-55^\circ\text{C} \leq T_A \leq +175^\circ\text{C}$			
Operating		$-55^\circ\text{C} \leq T_A \leq +200^\circ\text{C}$			
Storage		$-65^\circ\text{C} \leq T_A \leq +250^\circ\text{C}$			

MECHANICAL

TO-99 PACKAGE

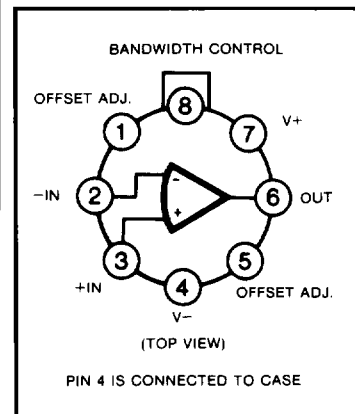


NOTE:
Leads in true position within 0.10° (.25mm) R @ MMC at seating plane.

Pin numbers shown for reference only.
Numbers may not be marked on package.

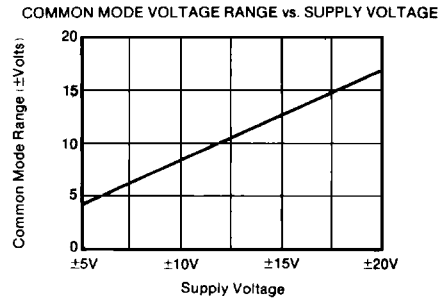
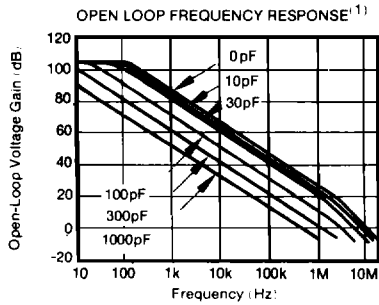
DIM	INCHES		MILLIMETERS	
	MIN	MAX	MIN	MAX
A	.335	.370	8.51	9.40
B	.305	.335	7.75	8.51
C	.165	.185	4.19	4.70
D	.018	.021	0.41	0.53
E	.010	.040	0.25	1.02
F	.010	.040	0.25	1.02
G	.200 BASIC		5.08 BASIC	
H	.028	.034	0.71	0.86
J	.029	.045	0.74	1.14
K	.500		12.7	
L	.110	.160	2.79	4.06
M	.45 ^B BASIC		45 ^B BASIC	
N	.095	.105	2.41	2.67

CONNECTION DIAGRAM

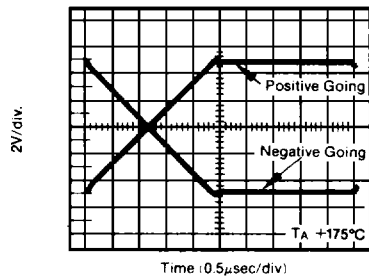


TYPICAL PERFORMANCE CURVES

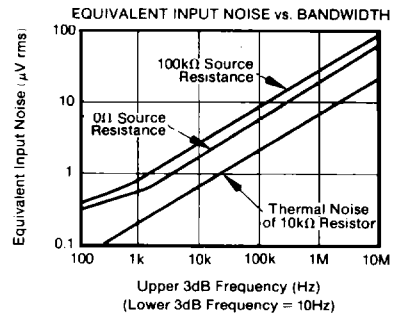
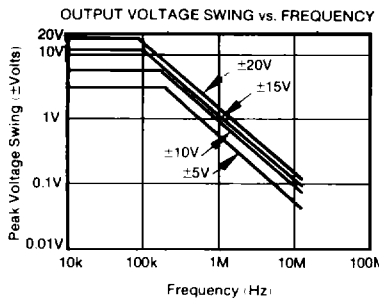
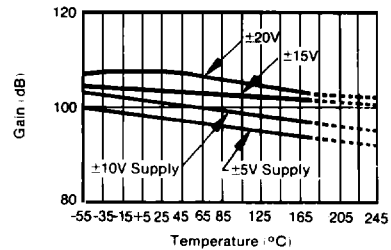
(at $\pm 15\text{VDC}$ and $T_A = +25^\circ\text{C}$ unless otherwise specified)



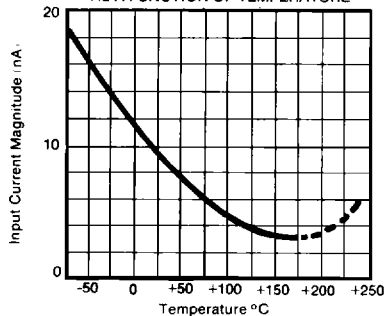
STEP RESPONSE IN FOLLOWER CONFIGURATION²⁾



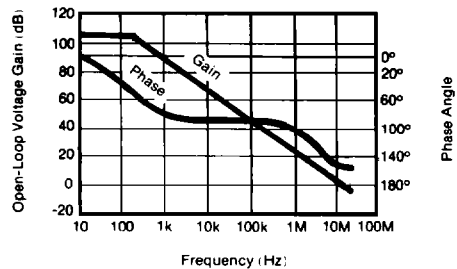
OPEN LOOP VOLTAGE GAIN vs. TEMPERATURE



INPUT BIAS CURRENT AND DIFFERENCE CURRENT AS A FUNCTION OF TEMPERATURE



OPEN-LOOP FREQUENCY AND PHASE RESPONSE



1. Capacitance values shown are compensation from pin 8 to common. Not required for stability. See Figure 1. 2. See Figure 3.

APPLICATIONS

BANDWIDTH COMPENSATION

The frequency response of the OPA11HT can be adjusted by use of an external compensation capacitor from pin 8 to common as shown in Figure 1. The open-loop frequency response curves illustrate the effect of various values of capacitance. The OPA11HT is stable at any gain level without the use of compensation, provided that stray wiring capacitance and/or load capacitance are not excessive, and that moderate values of feedback resistance are used ($R_{FB} \leq 10k\Omega$). A load capacitance of $\approx 50pF$ is desirable in all feedback configurations.

STABILITY

Because the OPA11HT is an extremely-fast amplifier with high gain, stray wiring capacitance and inductance in power supply leads can cause circuit oscillation. This can be prevented by proper circuit layout (all leads or patterns as short as possible) and by properly by passing the power supply lines to common at points close to the amplifier. In addition, it is recommended that the load be bypassed by a 50pF capacitor, see Figure 1.

OFFSET VOLTAGE AND ADJUSTMENT

Although the offset voltage of these amplifiers is only a few millivolts, it may in some cases be desirable to null this offset. This is done by use of a 100k Ω potentiometer as shown in Figure 2.

TEST CIRCUIT - DYNAMIC RESPONSE

The test circuit of Figure 3 is used for measurement of slew rate, settling time, rise time and overshoot. Both rise time and overshoot are measured for a small output signal ($V_{OUT} = \pm 100mV$). Slew rate and settling time are measured for a 10V, p-p, square wave.

VOLTAGE REGULATOR AT 200°C

In many applications, a regulated source of $\pm 15V$ is needed. A voltage regulator that typically will operate up to $+175^\circ C$ is shown in Figure 4. This regulator accepts $+16V$ to $+30V$ at its input and provides $+15V$ at 20mA at its output. A complementary version may be constructed to provide $-15V$ by using the OPA11HT with a 2N1711 transistor. Short-circuit protection should be added if required.

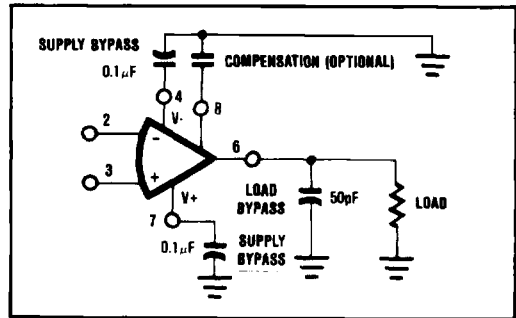


FIGURE 1. Compensated Amplifier with Supply Load Bypassing.

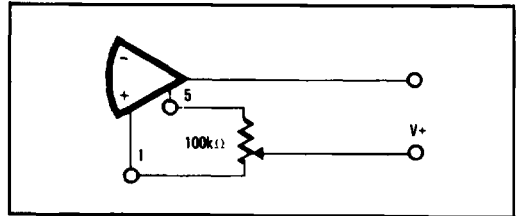


FIGURE 2. External Adjustment of Offset Voltage.

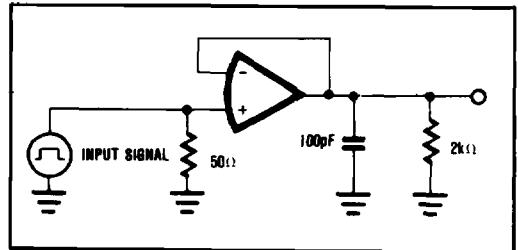


FIGURE 3. Dynamic Response Test Circuit.

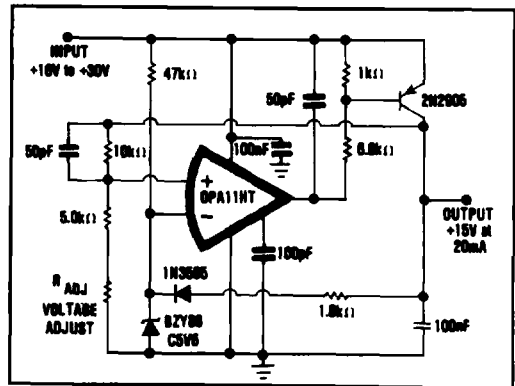


FIGURE 4. A +15V Voltage Regulator that will Operate at $+175^\circ C$.