16Mx64 DDR SDRAM

FEATURES

- DDR Data Rate = 200, 250, 266Mbps
- Package:
  - 219 Plastic Ball Grid Array (PBGA), 21 x 25mm
- 2.5V ±0.2V core power supply
- 2.5V I/O (SSTL_2 compatible)
- Differential clock inputs (CLK and CLK#)
- Commands entered on each positive CLK edge
- Internal pipelined double-data-rate (DDR) architecture; two data accesses per clock cycle
- Programmable Burst length: 2,4 or 8
- Bidirectional data strobe (DQS) transmitted/ received with data, i.e., source-synchronous data capture (one per byte)
- DQS edge-aligned with data for READs; center-aligned with data for WRITEs
- DLL to align DQ and DQS transitions with CLK
- Four internal banks for concurrent operation
- Two data mask (DM) pins for masking write data
- Programmable IOL/IOH option
- Auto precharge option
- Auto Refresh and Self Refresh Modes
- Commercial, Industrial and Military Temperature Ranges
- Organized as 16M x 64
- Weight: W3E16M64S-XBX - 2 grams typical

* This product is subject to change without notice.

BENEFITS

- 50% SPACE SAVINGS
- Reduced part count
- Reduced I/O count
  - 17% I/O Reduction
- Reduced trace lengths for lower parasitic capacitance
- Suitable for hi-reliability applications
- Laminate interposer for optimum TCE match
- Upgradeable to 32M x 64 density (W3E32M64S-XBX)
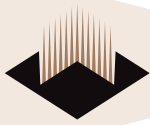
GENERAL DESCRIPTION

The 128MByte (1Gb) DDR SDRAM is a high-speed CMOS, dynamic random-access, memory using 4 chips containing 268,435,456 bits. Each chip is internally configured as a quad-bank DRAM. Each of the chip's 67,108,864-bit banks is organized as 8,192 rows by 512 columns by 16 bits.

The 128 MB DDR SDRAM uses a double data rate architecture to achieve high-speed operation. The double data rate architecture is essentially a 2n-prefetch architecture with an interface designed to transfer two data words per clock cycle at the I/O pins. A single read or write access for the 128MB DDR SDRAM effectively consists of a single 2n-bit wide, one-clock-cycle data tansfer at the internal DRAM core and two corresponding n-bit wide, one-half-clock-cycle data transfers at the I/O pins.

A bidirectional data strobe (DQS) is transmitted externally, along with data, for use in data capture at the receiver. DQS is a strobe transmitted by the DDR SDRAM during READs

	Monolithic Solution	Actual Size W3E16M64S-XBX	S A V I N G S
	11.9 11.9 11.9 11.9 ‹— 22.3 —› 66 TSOP 66 TSOP 66 TSOP 66 TSOP	21 25 WHITE ELECTRONIC DESIGNS W3E16M64S-XBX	
Area	4 x 265mm^2 = 1060mm^2	525mm^2	50%
I/O Count	4 x 66 pins = 264 pins	219 Balls	17%

W3E16M64S-XBX

FIGURE 1 – PIN CONFIGURATION

Top View

|	1	2	3	4	5	6	7	8	9	10	11	12	13	14	15	16
A		DQ13	DQ11	DQ9	V_{SS}	V_{SS}	A9	A10	A11	A8	V_{CCQ}	V_{CCQ}	DQ25	DQ27	DQ29	V_{SS}
B	DQ15	DQ2	DQ4	DQ6	V_{SS}	V_{SS}	A0	A7	A6	A1	V_{CC}	V_{CC}	DQ22	DQ20	DQ18	DQ31
C	DQ0	DQ12	DQ10	DQ8	V_{CC}	V_{CC}	A2	A5	A4	A3	V_{SS}	V_{SS}	DQ24	DQ26	DQ28	DQ16
D	DQ14	DQ3	DQ5	DQ7	V_{CCQ}	V_{CCQ}	A12	DNU	DNU	DNU	V_{SS}	V_{SS}	DQ23	DQ21	DQ19	DQ30
E	DQ1	DQML0	V_{CC}	DQMH0	DQSH3	DQSL0	DQSH0	BA0	BA1	DQSL1	DQSH1	VREF	DQML1	V_{SS}	NC	DQ17
F	CAS0#	WE0#	V_{CC}	CLK0	DQSL3							RAS1#	WE1#	V_{SS}	DQMH1	CLK1
G	CS0#	RAS0#	V_{CC}	CKE0	CLK0#							CAS1#	CS1#	V_{SS}	CLK1#	CKE1
H	V_{SS}	V_{SS}	V_{CC}	V_{CCQ}	V_{SS}							V_{CC}	V_{SS}	V_{SS}	V_{CCQ}	V_{CC}
J	V_{SS}	V_{SS}	V_{CC}	V_{CCQ}	V_{SS}							V_{CC}	V_{SS}	V_{SS}	V_{CCQ}	V_{CC}
K	CLK3#	CKE3	V_{CC}	CS3#	NC							CLK2#	CKE2	V_{SS}	RAS2#	CS2#
L	NC	CLK3	V_{CC}	CAS3#	RAS3#							DQSL2#	CLK2	V_{SS}	WE2#	CAS2#
M	DQ49	DQMH3	V_{CC}	WE3#	DQML3	NC	NC	NC	NC	NC	NC	NC	DQMH2	V_{SS}	DQML2	DQ33
N	DQ62	DQ50	DQ52	DQ54	NC	NC	NC	NC	NC	NC	NC	DQSH2	DQ39	DQ38	DQ35	DQ46
P	DQ48	DQ60	DQ57	DQ56	V_{SS}	V_{SS}	NC	NC	NC	NC	V_{CC}	V_{CC}	DQ41	DQ42	DQ44	DQ32
R	DQ63	DQ51	DQ53	DQ55	V_{CC}	V_{CC}	NC	NC	NC	NC	V_{SS}	V_{SS}	DQ37	DQ36	DQ34	DQ47
T	V_{SS}	DQ61	DQ59	DQ58	V_{CCQ}	V_{CCQ}	NC	NC	NC	NC	V_{SS}	V_{SS}	DQ40	DQ43	DQ45	V_{CC}

NOTE: DNU = Do Not Use; to be left unconnected for future upgrades.

NC = Not Connected Internally.

FIGURE 2 – FUNCTIONAL BLOCK DIAGRAM
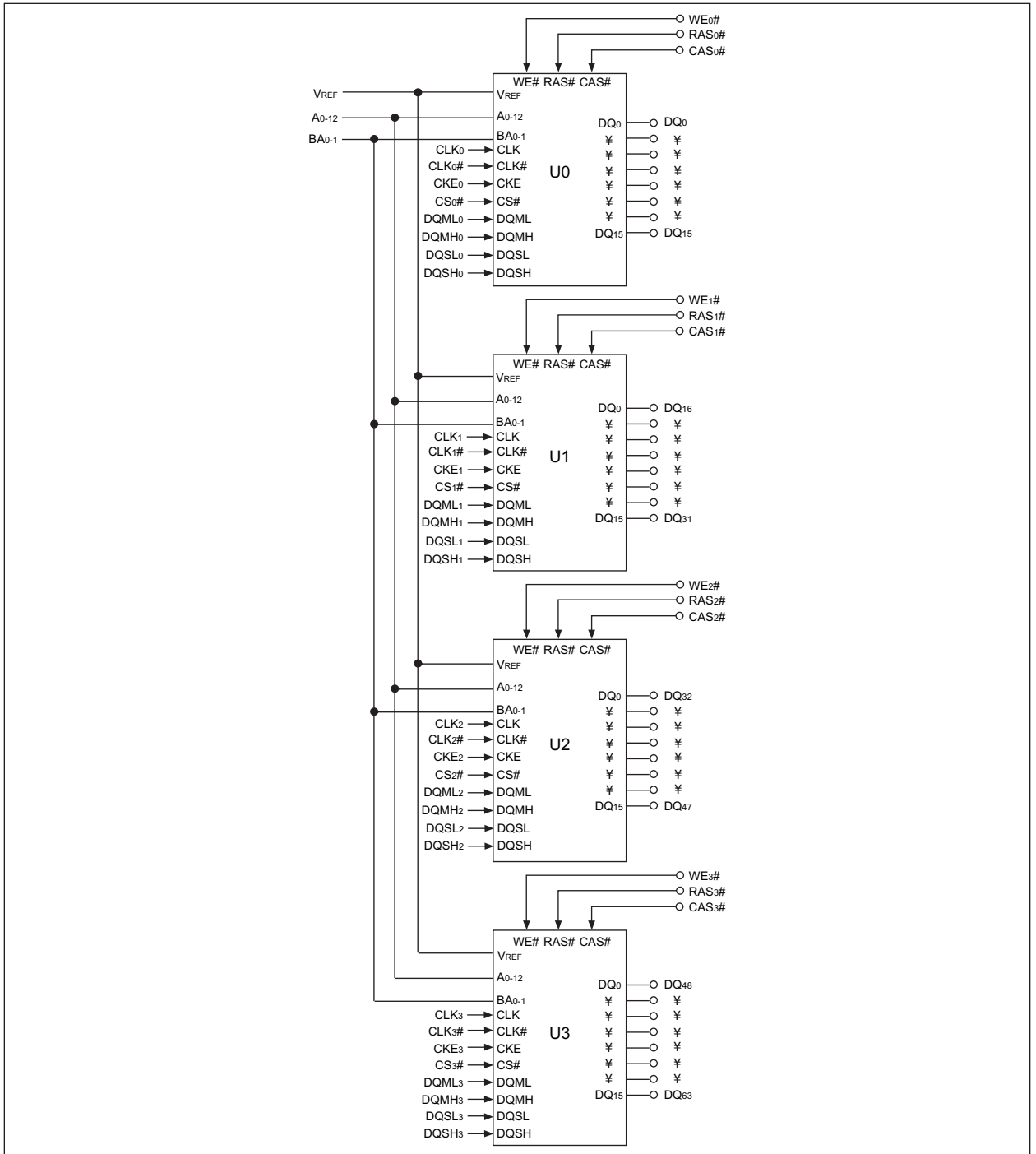
and by the memory contoller during WRITEs. DQS is edge-aligned with data for READs and center-aligned with data for WRITEs. Each chip has two data strobes, one for the lower byte and one for the upper byte.

The 128MB DDR SDRAM operates from a differential clock (CLK and CLK#); the crossing of CLK going HIGH and CLK# going LOW will be referred to as the positive edge of CLK. Commands (address and control signals) are registered at every positive edge of CLK. Input data is registered on both edges of DQS, and output data is referenced to both edges of DQS, as well as to both edges of CLK.

Read and write accesses to the DDR SDRAM are burst oriented; accesses start at a selected location and continue for a programmed number of locations in a programmed sequence. Accesses begin with the registration of an ACTIVE command, which is then followed by a READ or WRITE command. The address bits registered coincident with the ACTIVE command are used to select the bank and row to be accessed. The address bits registered coincident with the READ or WRITE command are used to select the bank and the starting column location for the burst access.

The DDR SDRAM provides for programmable READ or WRITE burst lengths of 2, 4, or 8 locations. An auto precharge function may be enabled to provide a self-timed row precharge that is initiated at the end of the burst access.

The pipelined, multibank architecture of DDR SDRAMs allows for concurrent operation, thereby providing high effective bandwidth by hiding row precharge and activation time.

An auto refresh mode is provided, along with a power-saving power-down mode.

FUNCTIONAL DESCRIPTION

Read and write accesses to the DDR SDRAM are burst oriented; accesses start at a selected location and continue for a programmed number of locations in a programmed sequence. Accesses begin with the registration of an ACTIVE command which is then followed by a READ or WRITE command. The address bits registered coincident with the ACTIVE command are used to select the bank and row to be accessed (BA0 and BA1 select the bank, A0-12 select the row). The address bits registered coincident with the READ or WRITE command are used to select the starting column location for the burst access.

Prior to normal operation, the SDRAM must be initialized. The following sections provide detailed information covering device initialization, register definition, command descriptions and device operation.

INITIALIZATION

DDR SDRAMs must be powered up and initialized in a predefined manner. Operational procedures other than those specified may result in undefined operation. Power must first be applied to V_{CC} and V_{CCQ} simultaneously, and then to V_{REF} (and to the system V_{TT}). V_{TT} must be applied after V_{CCQ} to avoid device latch-up, which may cause permanent damage to the device. V_{REF} can be applied any time after V_{CCQ} but is expected to be nominally coincident with V_{TT}. Except for CKE, inputs are not recognized as valid until after V_{REF} is applied. CKE is an SSTL_2 input but will detect an LVCMOS LOW level after V_{CC} is applied. Maintaining an LVCMOS LOW level on CKE during power-up is required to ensure that the DQ and DQS outputs will be in the High-Z state, where they will remain until driven in normal operation (by a read access). After all power supply and reference voltages are stable, and the clock is stable, the DDR SDRAM requires a $200 \propto s$ delay prior to applying an executable command.

Once the $200 \propto s$ delay has been satisfied, a DESELECT or NOP command should be applied, and CKE should be brought HIGH. Following the NOP command, a PRECHARGE ALL command should be applied. Next a LOAD MODE REGISTER command should be issued for the extended mode register (BA1 LOW and BA0 HIGH) to enable the DLL, followed by another LOAD MODE REGISTER command to the mode register (BA0/BA1 both LOW) to reset the DLL and to program the operating parameters. Two-hundred clock cycles are required between the DLL reset and any READ command. A PRECHARGE ALL command should then be applied, placing the device in the all banks idle state.

Once in the idle state, two AUTO REFRESH cycles must be performed (t_{RFC} must be satisfied.) Additionally, a LOAD MODE REGISTER command for the mode register with the reset DLL bit deactivated (i.e., to program operating parameters without resetting the DLL) is required. Following these requirements, the DDR SDRAM is ready for normal operation.
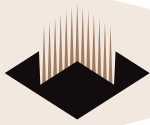
REGISTER DEFINITION

MODE REGISTER

The Mode Register is used to define the specific mode of operation of the DDR SDRAM. This definition includes the selection of a burst length, a burst type, a CAS latency, and an operating mode, as shown in Figure 3. The Mode Register is programmed via the MODE REGISTER SET command (with BA0 = 0 and BA1 = 0) and will retain the stored information until it is programmed again or the device loses power. (Except for bit A8 which is self clearing).

Reprogramming the mode register will not alter the contents of the memory, provided it is performed correctly. The Mode Register must be loaded (reloaded) when all banks are idle and no bursts are in progress, and the controller must wait the specified time before initiating the subsequent operation. Violating either of these requirements will result in unspecified operation.

Mode register bits A0-A2 specify the burst length, A3 specifies the type of burst (sequential or interleaved), A4-A6 specify the CAS latency, and A7-A12 specify the operating mode.

BURST LENGTH

Read and write accesses to the DDR SDRAM are burst oriented, with the burst length being programmable, as shown in Figure 3. The burst length determines the maximum number of column locations that can be accessed for a given READ or WRITE command. Burst lengths of 2, 4 or 8 locations are available for both the sequential and the interleaved burst types.

Reserved states should not be used, as unknown operation or incompatibility with future versions may result.

When a READ or WRITE command is issued, a block of columns equal to the burst length is effectively selected. All accesses for that burst take place within this block, meaning that the burst will wrap within the block if a boundary is reached. The block is uniquely selected by A1-Ai when the burst length is set to two; by A2-Ai when the burst length is set to four (where Ai is the most significant column address for a given configuration); and by A3-Ai when the burst length is set to eight. The remaining (least significant) address bit(s) is (are) used to select the starting location within the block. The programmed burst length applies to both READ and WRITE bursts.

BURST TYPE

Accesses within a given burst may be programmed to be either sequential or interleaved; this is referred to as the burst type and is selected via bit M3.

The ordering of accesses within a burst is determined by the burst length, the burst type and the starting column address, as shown in Table 1.

READ LATENCY

The READ latency is the delay, in clock cycles, between the registration of a READ command and the availability of the first bit of output data. The latency can be set to 2 or 2.5 clocks.

If a READ command is registered at clock edge n, and the latency is m clocks, the data will be available by clock edge $n+m$. Table 2 below indicates the operating frequencies at which each CAS latency setting can be used.

Reserved states should not be used as unknown operation or incompatibility with future versions may result.

TABLE 2 – CAS LATENCY

SPEED	ALLOWABLE OPERATING FREQUENCY (MHz)	
	CAS LATENCY = 2	CAS LATENCY = 2.5
-200	≤ 75	≤ 100
-250	≤ 100	≤ 125
-266	≤ 100	≤ 133

OPERATING MODE

The normal operating mode is selected by issuing a MODE REGISTER SET command with bits A7-A12 each set to zero, and bits A0-A6 set to the desired values. A DLL reset is initiated by issuing a MODE REGISTER SET command with bits A7 and A9-A12 each set to zero, bit A8 set to one, and bits A0-A6 set to the desired values. Although not required, JEDEC specifications recommend when a LOAD MODE REGISTER command is issued to reset the DLL, it should always be followed by a LOAD MODE REGISTER command to select normal operating mode.

All other combinations of values for A7-A12 are reserved for future use and/or test modes. Test modes and reserved states should not be used because unknown operation or incompatibility with future versions may result.
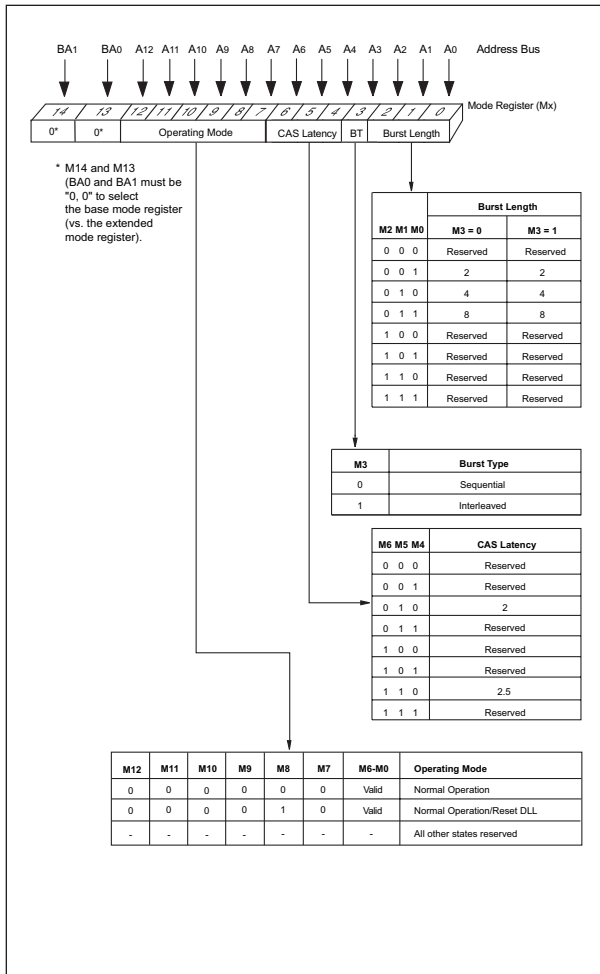
FIGURE 3 – MODE REGISTER DEFINITION

Address Bus: BA1 BA0 A12 A11 A10 A9 A8 A7 A6 A5 A4 A3 A2 A1 A0

Mode Register (Mx) bits: 14 13 12 11 10 9 8 7 6 5 4 3 2 1 0

| 0* | 0* | Operating Mode | CAS Latency | BT | Burst Length |

* M14 and M13 (BA0 and BA1 must be "0, 0" to select the base mode register (vs. the extended mode register).

M2 M1 M0	Burst Length M3 = 0	Burst Length M3 = 1
0 0 0	Reserved	Reserved
0 0 1	2	2
0 1 0	4	4
0 1 1	8	8
1 0 0	Reserved	Reserved
1 0 1	Reserved	Reserved
1 1 0	Reserved	Reserved
1 1 1	Reserved	Reserved

M3	Burst Type
0	Sequential
1	Interleaved

M6 M5 M4	CAS Latency
0 0 0	Reserved
0 0 1	Reserved
0 1 0	2
0 1 1	Reserved
1 0 0	Reserved
1 0 1	Reserved
1 1 0	2.5
1 1 1	Reserved

M12	M11	M10	M9	M8	M7	M6-M0	Operating Mode
0	0	0	0	0	0	Valid	Normal Operation
0	0	0	0	1	0	Valid	Normal Operation/Reset DLL
-	-	-	-	-	-	-	All other states reserved

TABLE 1 – BURST DEFINITION

Burst Length	Starting Column Address		Order of Accesses Within a Burst	
			Type = Sequential	Type = Interleaved
2		A0		
		0	0-1	0-1
		1	1-0	1-0
4	A1	A0		
	0	0	0-1-2-3	0-1-2-3
	0	1	1-2-3-0	1-0-3-2
	1	0	2-3-0-1	2-3-0-1
	1	1	3-0-1-2	3-2-1-0
8	A2	A1 A0		
	0	0 0	0-1-2-3-4-5-6-7	0-1-2-3-4-5-6-7
	0	0 1	1-2-3-4-5-6-7-0	1-0-3-2-5-4-7-6
	0	1 0	2-3-4-5-6-7-0-1	2-3-0-1-6-7-4-5
	0	1 1	3-4-5-6-7-0-1-2	3-2-1-0-7-6-5-4
	1	0 0	4-5-6-7-0-1-2-3	4-5-6-7-0-1-2-3
	1	0 1	5-6-7-0-1-2-3-4	5-4-7-6-1-0-3-2
	1	1 0	6-7-0-1-2-3-4-5	6-7-4-5-2-3-0-1
	1	1 1	7-0-1-2-3-4-5-6	7-6-5-4-3-2-1-0

NOTES:
1. For a burst length of two, A1-Ai select two-data-element block; A0 selects the starting column within the block.
2. For a burst length of four, A2-Ai select four-data-element block; A0-1 select the starting column within the block.
3. For a burst length of eight, A3-Ai select eight-data-element block; A0-2 select the starting column within the block.
4. Whenever a boundary of the block is reached within a given sequence above, the following access wraps within the block.

EXTENDED MODE REGISTER

The extended mode register controls functions beyond those controlled by the mode register; these additional functions are DLL enable/disable, output drive strength, and QFC#. These functions are controlled via the bits shown in Figure 5. The extended mode register is programmed via the LOAD MODE REGISTER command to the mode register (with BA0 = 1 and BA1 = 0) and will retain the stored information until it is programmed again or the device loses power. The enabling of the DLL should always be followed by a LOAD MODE REGISTER command to the mode register (BA0/BA1 both LOW) to reset the DLL.

The extended mode register must be loaded when all banks are idle and no bursts are in progress, and the controller must wait the specified time before initiating any subsequent operation. Violating either of these requirements could result in unspecified operation.
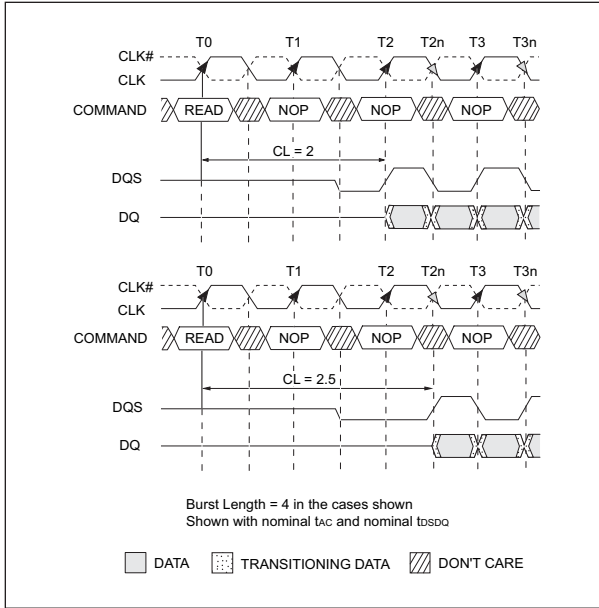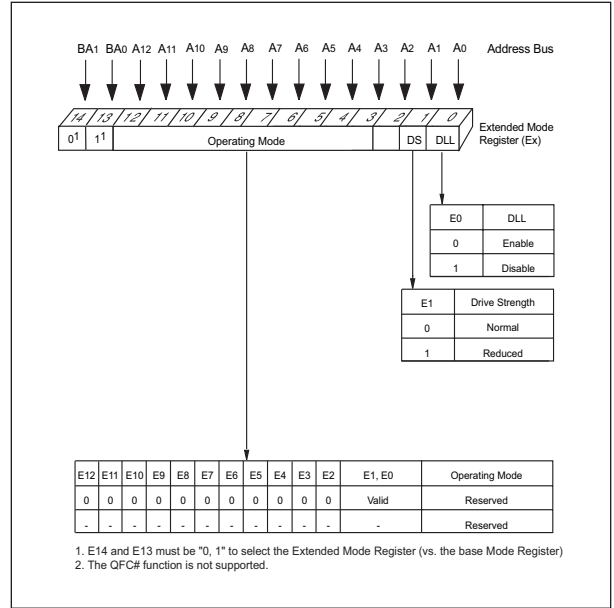
FIGURE 4 – CAS LATENCY

Burst Length = 4 in the cases shown
Shown with nominal t_{AC} and nominal t_{DSDQ}

DATA   TRANSITIONING DATA   DON'T CARE

FIGURE 5 – EXTENDED MODE REGISTER DEFINITION

E0	DLL
0	Enable
1	Disable

E1	Drive Strength
0	Normal
1	Reduced

E12	E11	E10	E9	E8	E7	E6	E5	E4	E3	E2	E1, E0	Operating Mode
0	0	0	0	0	0	0	0	0	0	0	Valid	Reserved
-	-	-	-	-	-	-	-	-	-	-	-	Reserved

1. E14 and E13 must be "0, 1" to select the Extended Mode Register (vs. the base Mode Register)
2. The QFC# function is not supported.

OUTPUT DRIVE STRENGTH

The normal full drive strength for all outputs are specified to be SSTL2, Class II. The DDR SDRAM supports an option for reduced drive. This option is intended for the support of the lighter load and/or point-to-point environments. The selection of the reduced drive strength will alter the DQs and DQSs from SSTL2, Class II drive strength to a reduced drive strength, which is approximately 54 percent of the SSTL2, Class II drive strength.

DLL ENABLE/DISABLE

The DLL must be enabled for normal operation. DLL enable is required during power-up initialization and upon returning to normal operation after having disabled the DLL for the purpose of debug or evaluation. (When the device exits self refresh mode, the DLL is enabled automatically.) Any time the DLL is enabled, 200 clock cycles must occur before a READ command can be issued.

COMMANDS

The Truth Table provides a quick reference of available commands. This is followed by a written description of each command.

DESELECT

The DESELECT function (CS# HiGH) prevents new commands from being executed by the DDR SDRAM. The SDRAM is effectively deselected. Operations already in progress are not affected.

NO OPERATION (NOP)

The NO OPERATION (NOP) command is used to perform a NOP to the selected DDR SDRAM (CS# is LOW). This prevents unwanted commands from being registered during idle or wait states. Operations already in progress are not affected.

LOAD MODE REGISTER

The Mode Registers are loaded via inputs A0-12. The LOAD MODE REGISTER command can only be issued when all banks are idle, and a subsequent executable command cannot be issued until t_{MRD} is met.

TRUTH TABLE – COMMANDS (NOTE 1)

NAME (FUNCTION)	CS#	RAS#	CAS#	WE#	ADDR
DESELECT (NOP) (9)	H	X	X	X	X
NO OPERATION (NOP) (9)	L	H	H	H	X
ACTIVE (Select bank and activate row) (3)	L	L	H	H	Bank/Row
READ (Select bank and column, and start READ burst) (4)	L	H	L	H	Bank/Col
WRITE (Select bank and column, and start WRITE burst) (4)	L	H	L	L	Bank/Col
BURST TERMINATE (8)	L	H	H	L	X
PRECHARGE (Deactivate row in bank or banks) (5)	L	L	H	L	Code
AUTO REFRESH or SELF REFRESH (Enter self refresh mode) (6, 7)	L	L	L	H	X
LOAD MODE REGISTER (2)	L	L	L	L	Op-Code

TRUTH TABLE – DM OPERATION

NAME (FUNCTION)	DM	DQs
WRITE ENABLE (10)	L	Valid
WRITE INHIBIT (10)	H	X

NOTES:
1. CKE is HIGH for all commands shown except SELF REFRESH.
2. A0-12 define the op-code to be written to the selected Mode Register. BA0, BA1 select either the mode register (0, 0) or the extended mode register (1, 0).
3. A0-12 provide row address, and BA0, BA1 provide bank address.
4. A0-8 provide column address; A10 HIGH enables the auto precharge feature (non persistent), while A10 LOW disables the auto precharge feature; BA0, BA1 provide bank address.
5. A10 LOW: BA0, BA1 determine the bank being precharged. A10 HIGH: All banks precharged and BA0, BA1 are "Don't Care."

6. This command is AUTO REFRESH if CKE is HIGH; SELF REFRESH if CKE is LOW.
7. Internal refresh counter controls row addressing; all inputs and I/Os are "Don't Care" except for CKE.
8. Applies only to read bursts with auto precharge disabled; this command is undefined (and should not be used) for READ bursts with auto precharge enabled and for WRITE bursts.
9. DESELECT and NOP are functionally interchangeable.
10. Used to mask write data; provided coincident with the corresponding data.

ACTIVE

The ACTIVE command is used to open (or activate) a row in a particular bank for a subsequent access. The value on the BA0, BA1 inputs selects the bank, and the address provided on inputs A0-12 selects the row. This row remains active (or open) for accesses until a PRECHARGE command is issued to that bank. A PRECHARGE command must be issued before opening a different row in the same bank.

READ

The READ command is used to initiate a burst read access to an active row. The value on the BA0, BA1 inputs selects the bank, and the address provided on inputs A0-8 selects the starting column location. The value on input A10 determines whether or not AUTO PRECHARGE is used. If AUTO PRECHARGE is selected, the row being accessed will be precharged at the end of the READ burst; if AUTO PRECHARGE is not selected, the row will remain open for subsequent accesses.

WRITE

The WRITE command is used to initiate a burst write access to an active row. The value on the BA0, BA1 inputs selects the bank, and the address provided on inputs A0-8 selects the starting column location. The value on input A10 determines whether or not AUTO PRECHARGE is used. If AUTO PRECHARGE is selected, the row being accessed will be precharged at the end of the WRITE burst; if AUTO PRECHARGE is not selected, the row will remain open for subsequent accesses. Input data appearing on the D/Qs is written to the memory array subject to the DQM input logic level appearing coincident with the data. If a given DQM signal is registered LOW, the corresponding data will be written to memory; if the DQM signal is registered HIGH, the corresponding data inputs will be ignored, and a WRITE will not be executed to that byte/column location.

PRECHARGE

The PRECHARGE command is used to deactivate the open row in a particular bank or the open row in all banks. The bank(s) will be available for a subsequent row access a specified time (tRP) after the PRECHARGE command is

issued. Except in the case of concurrent auto precharge, where a READ or WRITE command to a different bank is allowed as long as it does not interrupt the data transfer in the current bank and does not violate any other timing parameters. Input A10 determines whether one or all banks are to be precharged, and in the case where only one bank is to be precharged, inputs BA0, BA1 select the bank. Otherwise BA0, BA1 are treated as "Don't Care." Once a bank has been precharged, it is in the idle state and must be activated prior to any READ or WRITE commands being issued to that bank. A PRECHARGE command will be treated as a NOP if there is no open row in that bank (idle state), or if the previously open row is already in the process of precharging.

AUTO PRECHARGE

AUTO PRECHARGE is a feature which performs the same individual-bank PRECHARGE function described above, but without requiring an explicit command. This is accomplished by using A10 to enable AUTO PRECHARGE in conjunction with a specific READ or WRITE command. A precharge of the bank/row that is addressed with the READ or WRITE command is automatically performed upon completion of the READ or WRITE burst. AUTO PRECHARGE is nonpersistent in that it is either enabled or disabled for each individual READ or WRITE command. The device supports concurrent auto precharge if the command to the other bank does not interrupt the data transfer to the current bank.

AUTO PRECHARGE ensures that the precharge is initiated at the earliest valid stage within a burst. This "earliest valid stage" is determined as if an explicit precharge command was issued at the earliest possible time, without violating t_{RAS} (MIN). The user must not issue another command to the same bank until the precharge time (t_{RP}) is completed. This is determined as if an explicit PRECHARGE command was issued at the earliest possible time, without violating t_{RAS} (MIN).

BURST TERMINATE

The BURST TERMINATE command is used to truncate READ bursts (with auto precharge disabled). The most recently registered READ command prior to the BURST TERMINATE command will be truncated. The open page which the READ burst was terminated from remains open.

AUTO REFRESH

AUTO REFRESH is used during normal operation of the DDR SDRAM and is analogous to CAS#-BEFORE-RAS#

(CBR) REFRESH in conventional DRAMs. This command is nonpersistent, so it must be issued each time a refresh is required.

The addressing is generated by the internal refresh controller. This makes the address bits "Don't Care" during an AUTO REFRESH command. Each DDR SDRAM requires AUTO REFRESH cycles at an average interval of $7.8125 \propto s$ (maximum).

To allow for improved efficiency in scheduling and switching between tasks, some flexibility in the absolute refresh interval is provided. A maximum of eight AUTO REFRESH commands can be posted to any given DDR SDRAM, meaning that the maximum absolute interval between any AUTO REFRESH command and the next AUTO REFRESH command is $9 \times 7.8125 \propto s$ ($70.3 \propto s$). This maximum absolute interval is to allow future support for DLL updates internal to the DDR SDRAM to be restricted to AUTO REFRESH cycles, without allowing excessive drift in t_{AC} between updates.

Although not a JEDEC requirement, to provide for future functionality features, CKE must be active (High) during the AUTO REFRESH period. The AUTO REFRESH period begins when the AUTO REFRESH command is registered and ends t_{RFC} later.
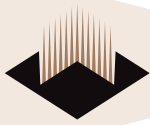
SELF REFRESH*

The SELF REFRESH command can be used to retain data in the DDR SDRAM, even if the rest of the system is powered down. When in the self refresh mode, the DDR SDRAM retains data without external clocking. The SELF REFRESH command is initiated like an AUTO REFRESH command except CKE is disabled (LOW). The DLL is automatically disabled upon entering SELF REFRESH and is automatically enabled upon exiting SELF REFRESH (200 clock cycles must then occur before a READ command can be issued). Input signals except CKE are "Don't Care" during SELF REFRESH.

The procedure for exiting self refresh requires a sequence of commands. First, CLK must be stable prior to CKE going back HIGH. Once CKE is HIGH, the DDR SDRAM must have NOP commands issued for t_{XSNR}, because time is required for the completion of any internal refresh in progress.

A simple algorithm for meeting both refresh and DLL requirements is to apply NOPs for 200 clock cycles before applying any other command.

* Self refresh available in commercial and industrial temperatures only.

ABSOLUTE MAXIMUM RATINGS

Parameter		Unit
Voltage on V_{CC}, V_{CCQ} Supply relative to Vss	-1 to 3.6	V
Voltage on I/O pins relative to Vss	-1 to 3.6	V
Operating Temperature T_A (Mil)	-55 to +125	°C
Operating Temperature T_A (Ind)	-40 to +85	°C
Storage Temperature, Plastic	-55 to +150	°C

NOTE:
Stress greater than those listed under "Absolute Maximum Ratings" may cause permanent damage to the device. This is a stress rating only and functional operation of the device at these or any other conditions greater than those indicated in the operational sections of this specification is not implied. Exposure to absolute maximum rating conditions for extended periods may affect reliability.

CAPACITANCE (NOTE 13)

Parameter	Symbol	Max	Unit
Input Capacitance: CLK	CI1	8	pF
Addresses, BA0-1 Input Capacitance	CA	30	pF
Input Capacitance: All other input-only pins	CI2	8	pF
Input/Output Capacitance: I/Os	CIO	12	pF

BGA THERMAL RESISTANCE

Description	Symbol	Max	Units	Notes
Junction to Ambient (No Airflow)	Theta JA	14.5	°C/W	1
Junction to Ball	Theta JB	10.0	°C/W	1
Junction to Case (Top)	Theta JC	5.4	°C/W	1

NOTE 1:
Refer to PBGA Thermal Resistance Correlation application note at www.wedc.com in the application notes section for modeling conditions.

DC ELECTRICAL CHARACTERISTICS AND OPERATING CONDITIONS (NOTES 1, 6)

V_{CC} = +2.5V ±0.2V; -55°C ≤ T_A ≤ +125°C

Parameter/Condition	Symbol	Min	Max	Units
Supply Voltage	V_{CC}	2.3	2.7	V
I/O Supply Voltage	V_{CCQ}	2.3	2.7	V
Input High Voltage: Logic 1; All inputs (21)	V_{IH}	V_{REF} - 0.04	V_{REF} + 0.04	V
Input Low Voltage: Logic 0; All inputs (21)	V_{IL}	-0.3	V_{REF} - 0.15	V
Input Leakage Current: Any input 0V ≤ V_{IN} ≤ V_{CC} (All other pins not under test = 0V)	II	-2	2	µA
Input Leakage Address Current (All other pins not under test = 0V)	I_I	-8	8	µA
Output Leakage Current: I/Os are disabled; 0V ≤ V_{OUT} ≤ V_{CC}	I_{OZ}	-5	5	µA
Output Levels: Full drive option High Current (V_{OUT} = V_{CCQ} - 0.373V, minimum V_{REF}, minimum V_{TT}) Low Current (V_{OUT} = 0.373V, maximum V_{REF}, maximum V_{TT})	I_{OH}	-12	—	mA
	I_{OL}	12	—	mA
Output Levels: Reduced drive option High Current (V_{OUT} = V_{CCQ} - 0.763V, minimum V_{REF}, minimum V_{TT}) Low Current (V_{OUT} = 0.763V, maximum V_{REF}, maximum V_{TT})	I_{OHR}	-9	—	mA
	I_{OLR}	9	—	mA
I/O Reference Voltage	V_{REF}	0.49 x V_{CCQ}	0.51 x V_{CCQ}	V
I/O Termination Voltage	V_{TT}	V_{REF} - 0.04	V_{REF} + 0.04	V

Icc SPECIFICATIONS AND CONDITIONS (NOTES 1-5, 10, 12, 14)

V_{CC} = +2.5V ±0.2V; -55°C ≤ T_A ≤ +125°C

Parameter/Condition		Symbol	Max		Units
			250Mbps 266Mbps	200Mbps	
OPERATING CURRENT: One bank; Active-Precharge; t_{RC} = t_{RC} (MIN); t_{CK} = t_{CK} (MIN); DQ, DM, and DQS inputs changing once per clock cyle; Address and control inputs changing once every two clock cycles; (22, 48)		I_{CC0}	500	480	mA
OPERATING CURRENT: One bank; Active-Read-Precharge; Burst = 2; t_{RC} = t_{RC} (MIN); t_{CK} = t_{CK} (MIN); I_{OUT} = 0mA; Address and control inputs changing once per clock cycle (22, 48)		I_{CC1}	680	620	mA
PRECHARGE POWER-DOWN STANDBY CURRENT: All banks idle; Power-down mode; t_{CK} = t_{CK} (MIN); CKE = LOW; (23, 32, 50)		I_{CC2P}	16	16	mA
IDLE STANDBY CURRENT: CS = HIGH; All banks idle; t_{CK} = t_{CK} (MIN); CKE = HIGH; Address and other control inputs changing once per clock cycle. V_{IN} = V_{REF} for DQ, DQS, and DM (51)		I_{CC2F}	180	180	mA
ACTIVE POWER-DOWN STANDBY CURRENT: One bank active; Power-down mode; t_{CK} = t_{CK} (MIN); CKE = LOW (23, 32, 50)		I_{CC3P}	120	120	mA
ACTIVE STANDBY CURRENT: CS = HIGH; CKE = HIGH; One bank; Active-Precharge; t_{RC} = t_{RAS} (MAX); t_{CK} = t_{CK} (MIN); DQ, DM, and DQS inputs changing twice per clock cycle; Address and other control inputs changing once per clock cycle (22)		I_{CC3N}	200	200	mA
OPERATING CURRENT: Burst = 2; Reads; Continuous burst; One bank active; Address and control inputs changing once per clock cycle; t_{CK} = t_{CK} (MIN); I_{OUT} = 0mA (22, 48)		I_{CC4R}	740	740	mA
OPERATING CURRENT: Burst = 2; Writes; Continuous burst; One bank active; Address and control inputs changing once per clock cycle; t_{CK} = t_{CK} (MIN); DQ, DM, and DQS inputs changing twice per clock cycle (22)		I_{CC4W}	640	640	mA
AUTO REFRESH CURRENT	t_{RC} = t_{RC} (MIN) (27, 50)	I_{CC5}	980	980	mA
	t_{RC} = 7.8125µs (27, 50)	I_{CC5A}	24	24	mA
SELF REFRESH CURRENT: CKE ≤ 0.2V	Standard (11)	I_{CC6}	16	16	mA
OPERATING CURRENT: Four bank interleaving READs (BL=4) with auto precharge, t_{RC} = t_{RC} (MIN); t_{CK} = t_{CK} (MIN); Address and control inputs change only during Active READ or WRITE commands. (22, 49)		I_{CC7}	1600	1600	mA

WHITE ELECTRONIC DESIGNS

## ELECTRICAL CHARACTERISTICS AND RECOMMENDED AC OPERATING CHARACTERISTICS

(Notes 1-5, 14-17, 33)

Parameter		Symbol	266 Mbps CL 2.5 200 Mbps CL 2		250 Mbps CL2.5 200 Mbps CL2		200 Mbps CL2.5 150 Mbps CL2		Units
			Min	Max	Min	Max	Min	Max	
Access window of DQs from CLK/CLK#		t_{AC}	-0.75	+0.75	-0.8	+0.8	-0.8	+0.8	ns
CLK high-level width (30)		t_{CH}	0.45	0.55	0.45	0.55	0.45	0.55	t_{CK}
CLK low-level width (30)		t_{CL}	0.45	0.55	0.45	0.55	0.45	0.55	t_{CK}
Clock cycle time	CL = 2.5 (45, 52)	t_{CK} (2.5)	7.5	13	8	13	10	13	ns
	CL = 2 (45, 52)	t_{CK} (2)	10	13	10	13	13	15	ns
DQ and DM input hold time relative to DQS (26, 31)		t_{DH}	0.5		0.6		0.6		ns
DQ and DM input setup time relative to DQS (26, 31)		t_{DS}	0.5		0.6		0.6		ns
DQ and DM input pulse width (for each input) (31)		t_{DIPW}	1.75		2		2		ns
Access window of DQS from CLK/CLK		t_{DQSCK}	-0.75	+0.75	-0.8	+0.8	-0.8	+0.8	ns
DQS input high pulse width		t_{DQSH}	0.35		0.35		0.35		t_{CK}
DQS input low pulse width		t_{DQSL}	0.35		0.35		0.35		t_{CK}
DQS-DQ skew, DQS to last DQ valid, per group, per access (25, 26)		t_{DQSQ}		0.5		0.6		0.6	ns
Write command to first DQS latching transition		t_{DQSS}	0.75	1.25	0.75	1.25	0.75	1.25	t_{CK}
DQS falling edge to CLK rising - setup time		t_{DSS}	0.2		0.2		0.2		t_{CK}
DQS falling edge from CLK rising - hold time		t_{DSH}	0.2		0.2		0.2		t_{CK}
Half clock period (34)		t_{HP}	t_{CH},t_{CL}		t_{CH},t_{CL}		t_{CH},t_{CL}		ns
Data-out high-impedance window from CLK/CLK (18, 42)		t_{HZ}		+0.75		+0.8		+0.8	ns
Data-out low-impedance window from CLK/CLK (18, 43)		t_{LZ}	-0.75		-0.8		-0.8		ns
Address and control input hold time (fast slew rate) (14)		t_{IH_F}	0.90		1.1		1.1		ns
Address and control input setup time (fast slew rate) (14)		t_{IS_F}	0.90		1.1		1.1		ns
Address and control input hold time (slow slew rate) (14)		t_{IH_S}	1		1.1		1.1		ns
Address and control input setup time (slow slew rate) (14)		t_{IS_S}	1		1.1		1.1		ns
LOAD MODE REGISTER command cycle time		t_{MRD}	15		16		16		ns
DQ-DQS hold, DQS to first DQ to go non-valid, per access (25, 26)		t_{QH}	$t_{HP}-t_{QHS}$		$t_{HP}-t_{QHS}$		$t_{HP}-t_{QHS}$		ns
Data hold skew factor		t_{QHS}		0.75		1		1	ns
ACTIVE to PRECHARGE command (35)		t_{RAS}	40	120,000	40	120,000	40	120,000	ns
ACTIVE to READ with Auto precharge command (46)		t_{RAP}	20		20		20		ns
ACTIVE to ACTIVE/AUTO REFRESH command period		t_{RC}	65		70		70		ns
AUTO REFRESH command period (50)		t_{RFC}	75		80		80		ns
ACTIVE to READ or WRITE delay		t_{RCD}	20		20		20		ns
PRECHARGE command period		t_{RP}	20		20		20		ns
DQS read preamble (42)		t_{RPRE}	0.9	1.1	0.9	1.1	0.9	1.1	t_{CK}
DQS read postamble		t_{RPST}	0.4	0.6	0.4	0.6	0.4	0.6	t_{CK}
ACTIVE bank a to ACTIVE bank b command		t_{RRD}	15		15		15		ns
DQS write preamble		t_{WPRE}	0.25		0.25		0.25		t_{CK}
DQS write preamble setup time (20, 21)		t_{WPRES}	0		0		0		ns
DQS write postamble (19)		t_{WPST}	0.4	0.6	0.4	0.6	0.4	0.6	t_{CK}
Write recovery time		t_{WR}	15		15		15		ns
Internal WRITE to READ command delay		t_{WTR}	1		1		1		t_{CK}
Data valid output window (25)		na	$t_{QH} - t_{DQSQ}$		$t_{QH} - t_{DQSQ}$		$t_{QH} - t_{DQSQ}$		ns
REFRESH to REFRESH command interval (23)		t_{REFC}		70.3		70.3		70.3	µs
Average periodic refresh interval (23)		t_{REFI}		7.8		7.8		7.8	µs
Terminating voltage delay to VDD		t_{VTD}	0		0		0		ns
Exit SELF REFRESH to non-READ command		t_{XSNR}	75		80		80		ns
Exit SELF REFRESH to READ command		t_{XSRD}	200		200		200		t_{CK}

NOTES:

1. All voltages referenced to V_{SS}.

2. Tests for AC timing, I_{CC}, and electrical AC and DC characteristics may be conducted at nominal reference/supply voltage levels, but the related specifications and device operation are guaranteed for the full voltage range specified.

3. Outputs measured with equivalent load:
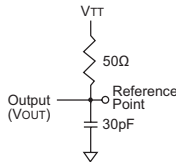
4. AC timing and I_{CC} tests may use a V_{IL}-to-V_{IH} swing of up to 1.5V in the test environment, but input timing is still referenced to V_{REF} (or to the crossing point for CLK/CLK#), and parameter specifications are guaranteed for the specified AC input levels under normal use conditions. The minimum slew rate for the input signals used to test the device is 1V/ns in the range between $V_{IL}(AC)$ and $V_{IH}(AC)$.

5. The AC and DC input level specifications are as defined in the SSTL_2 Standard (i.e., the receiver will effectively switch as a result of the signal crossing the AC input level, and will remain in that state as long as the signal does not ring back above [below] the DC input LOW [HIGH] level).

6. V_{REF} is expected to equal $V_{CCQ/2}$ of the transmitting device and to track variations in the DC level of the same. Peak-to-peak noise (noncommon mode) on V_{REF} may not exceed ±2 percent of the DC value. Thus, from $V_{CCQ/2}$, V_{REF} is allowed ±25mV for DC error and an additional ±25mV for AC noise. This measurement is to be taken at the nearest V_{REF} by-pass capacitor.

7. V_{TT} is not applied directly to the device. V_{TT} is a system supply for signal termination resistors, is expected to be set equal to V_{REF} and must track variations in the DC level of V_{REF}.

8. V_{ID} is the magnitude of the difference between the input level on CLK and the input level on CLK#.

9. The value of V_{IX} and V_{MP} are expected to equal $V_{CCQ/2}$ of the transmitting device and must track variations in the DC level of the same.

10. I_{CC} is dependent on output loading and cycle rates. Specified values are obtained with minimum cycle time with the outputs open.

11. Enables on-chip refresh and address counters.

12. I_{CC} specifications are tested after the device is properly initialized, and is averaged at the defined cycle rate.

13. This parameter is not tested but guaranteed by design. $t_A = 25°C$, f = 1 MHz

14. Command/Address input slew rate = 0.5V/ns. For 266 MHz with slew rates 1V/ns and faster, tIS and tIH are reduced to 900ps. If the slew rate is less than 0.5V/ns,

timing must be derated: tIS has an additional 50ps per each 100mV/ns reduction in slew rate from the 500mV/ns. tIH has 0ps added, that is, it remains constant. If the slew rate exceeds 4.5V/ns, functionality is uncertain.

15. The CLK/CLK# input reference level (for timing referenced to CLK/CLK#) is the point at which CLK and CLK# cross; the input reference level for signals other than CLK/CLK# is V_{REF}.

16. Inputs are not recognized as valid until V_{REF} stabilizes. Exception: during the period before V_{REF} stabilizes, CKE ≤ 0.3 x V_{CCQ} is recognized as LOW.

17. The output timing reference level, as measured at the timing reference point indicated in Note 3, is V_{TT}.

18. t_{HZ} and t_{LZ} transitions occur in the same access time windows as valid data transitions. These parameters are not referenced to a specific voltage level, but specify when the device output is no longer driving (HZ) or begins driving (LZ).

19. The maximum limit for this parameter is not a device limit. The device will operate with a greater value for this parameter, but system performance (bus turnaround) will degrade accordingly.

20. This is not a device limit. The device will operate with a negative value, but system performance could be degraded due to bus turnaround.

21. It is recommended that DQS be valid (HIGH or LOW) on or before the WRITE command. The case shown (DQS going from High-Z to logic LOW) applies when no WRITEs were previously in progress on the bus. If a previous WRITE was in progress, DQS could be HIGH during this time, depending on t_{DQSS}.

22. MIN (t_{RC} or t_{RFC}) for I_{CC} measurements is the smallest multiple of t_{CK} that meets the minimum absolute value for the respective parameter. t_{RAS} (MAX) for I_{CC} measurements is the largest multiple of t_{CK} that meets the maximum absolute value for t_{RAS}.

23. The refresh period 64ms. This equates to an average refresh rate of 7.8125µs. However, an AUTO REFRESH command must be asserted at least once every 70.3µs; burst refreshing or posting by the DRAM controller greater than eight refresh cycles is not allowed.

24. The I/O capacitance per DQS and DQ byte/group will not differ by more than this maximum amount for any given device.

25. The valid data window is derived by achieving other specifications - t_{HP} ($t_{CK/2}$), t_{DQSQ}, and t_{QH} ($t_{QH} = t_{HP} - t_{QHS}$). The data valid window derates directly porportional with the clock duty cycle and a practical data valid window can be derived. The clock is allowed a maximum duty cycle variation of 45/55. Functionality is uncertain when operating beyond a 45/55 ratio. The data valid window derating curves are provided below for duty cycles ranging between 50/50 and 45/55.

26. Referenced to each output group: LDQS with DQ0-DQ7; and UDQS with DQ8-DQ15 of each chip.

27. This limit is actually a nominal value and does not result in a fail value. CKE is HIGH during REFRESH command period (t_{RFC} [MIN]) else CKE is LOW (i.e., during standby).
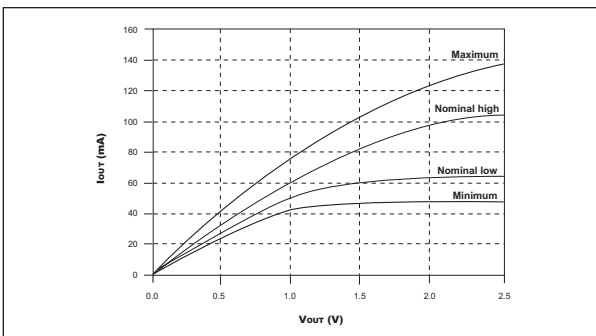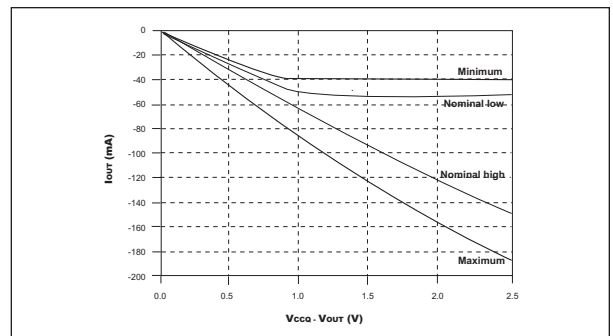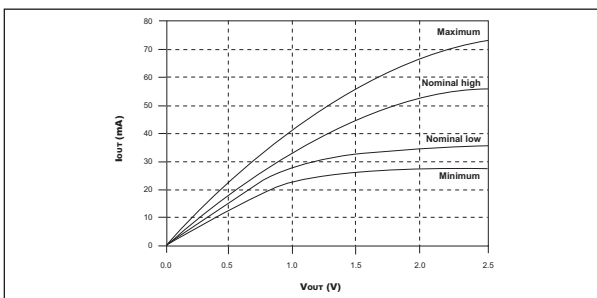
FIGURE A – PULL-DOWN CHARACTERISTICS
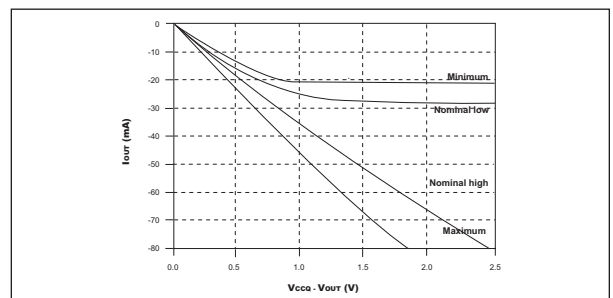

FIGURE B – PULL-UP CHARACTERISTICS

28. To maintain a valid level, the transitioning edge of the input must:
    a) Sustain a constant slew rate from the current AC level through to the target AC level, $V_{IL}(AC)$ or $V_{IH}(AC)$.
    b) Reach at least the target AC level.
    c) After the AC target level is reached, continue to maintain at least the target DC level, $V_{IL}(DC)$ or $V_{IH}(DC)$.
29. The Input capacitance per pin group will not differ by more than this maximum amount for any given device.
30. CLK and CLK# input slew rate must be ≥ 1V/ns (≥2V/ns differentially).
31. DQ and DM input slew rates must not deviate from DQS by more than 10%. If the DQ/DM/DQS slew rate is less than 0.5V/ns, timing must be derated: 50ps must be added to t_{DS} and t_{DH} for each 100mV/ns reduction in slew rate. If slew rate exceeds 4V/ns, functionality is uncertain.
32. V_{CC} must not vary more than 4% if CKE is not active while any bank is active.
33. The clock is allowed up to ±150ps of jitter. Each timing parameter is allowed to vary by the same amount.
34. t_{HP} min is the lesser of t_{CL} minimum and t_{CH} minimum actually applied to the device CLK and CLK# inputs, collectively during bank active.
35. READs and WRITEs with auto precharge are not allowed to be issued until $t_{RAS}(MIN)$ can be satisfied prior to the internal precharge command being issued.
36. Any positive glitch must be less than 1/3 of the clock and not more than +400mV or 2.9 volts, whichever is less. Any negative glitch must be less than 1/3 of the clock cycle and not exceed either -300mV or 2.2 volts, whichever is more positive.
37. Normal Output Drive Curves:
    a) The full variation in driver pull-down current from minimum to maximum process, temperature and voltage will lie within the outer bounding lines of the V-I curve of Figure A.
    b) The variation in driver pull-down current within nominal limits of voltage and temperature is expected, but not guaranteed, to lie within the inner bounding lines of the V-I curve of Figure A.
    c) The full variation in driver pull-up current from minimum to maximum process, temperature and voltage will lie within the outer bounding lines of the V-I curve of Figure B.
    d) The variation in driver pull-up current within nominal limits of voltage and temperature is expected, but not guaranteed, to lie within the inner bounding lines of the V-I curve of Figure B.
    e) The full variation in the ratio of the maximum to minimum pull-up and pull-down current should be between .71 and 1.4, for device drain-to-source voltages from 0.1V to 1.0 Volt, and at the same voltage and temperature.
    f) The full variation in the ratio of the nominal pull-up to pull-down current should be unity ±10%, for device drain-to-source voltages from 0.1V to 1.0 Volt.
38. Reduced Output Drive Curves:
    a) The full variation in driver pull-down current from minimum to maximum process, temperature and voltage will lie within the outer bounding lines of the V-I curve of Figure C.

b) The variation in driver pull-down current within nominal limits of voltage and temperature is expected, but not guaranteed, to lie within the inner bounding lines of the V-I curve of Figure C.
c) The full variation in driver pull-up current from minimum to maximum process, temperature and voltage will lie within the outer bounding lines of the V-I curve of Figure D.
d) The variation in driver pull-up current within nominal limits of voltage and temperature is expected, but not guaranteed, to lie within the inner bounding lines of the V-I curve of Figure D.
e) The full variation in the ratio of the maximum to minimum pull-up and pull-down current should be between .71 and 1.4, for device drain-to-source voltages from 0.1V to 1.0 V, and at the same voltage and temperature.
f) The full variation in the ratio of the nominal pull-up to pull-down current should be unity ±10%, for device drain-to-source voltages from 0.1V to 1.0 V.
39. The voltage levels used are derived from a minimum V_{CC} level and the referenced test load. In practice, the voltage levels obtained from a properly terminated bus will provide significantly different voltage values.
40. V_{IH} overshoot: $V_{IH}(MAX) = V_{CCQ}+1.5V$ for a pulse width ≤ 3ns and the pulse width can not be greater than 1/3 of the cycle rate.
41. V_{CC} and V_{CCQ} must track each other.
42. This maximum value is derived from the referenced test load. In practice, the values obtained in a typical terminated design may reflect up to 310ps less for $t_{HZ}(MAX)$ and the last DVW. $t_{HZ}(MAX)$ will prevail over $t_{DQSCK}(MAX) + t_{RPST}(MAX)$ condition. $t_{LZ}(MIN)$ will prevail over $t_{DQSCK}(MIN) + t_{RPRE}(MAX)$ condition.
43. For slew rates greater than 1V/ns the (LZ) transition will start about 310ps earlier.
44. During initialization, V_{CCQ}, V_{TT}, and V_{REF} must be equal to or less than $V_{CC} + 0.3V$. Alternatively, V_{TT} may be 1.35V maximum during power up, even if V_{CC}/V_{CCQ} are 0 volts, provided a minimum of 42 ohms of series resistance is used between the V_{TT} supply and the input pin.
45. The current part operates below the slowest JEDEC operating frequency of 83 MHz. As such, future die may not reflect this option.
46. Reserved for future use.
47. Reserved for future use.
48. Random addressing changing 50% of data changing at every transfer.
49. Random addressing changing 100% of data changing at every transfer.
50. CKE must be active (high) during the entire time a refresh command is executed. That is, from the time the AUTO REFRESH command is registered, CKE must be active at each rising clock edge, until t_{RFC} has been satisfied.
51. I_{CC2N} specifies the DQ, DQS, and DM to be driven to a valid high or low logic level. I_{CC2Q} is similar to I_{CC2F} except I_{CC2Q} specifies the address and control inputs to remain stable. Although I_{CC2F}, I_{CC2N}, and I_{CC2Q} are similar, I_{CC2Q} is "worst case."
52. Whenever the operating frequency is altered, not including jitter, the DLL is required to be reset. This is followed by 200 clock cycles before any READ command.
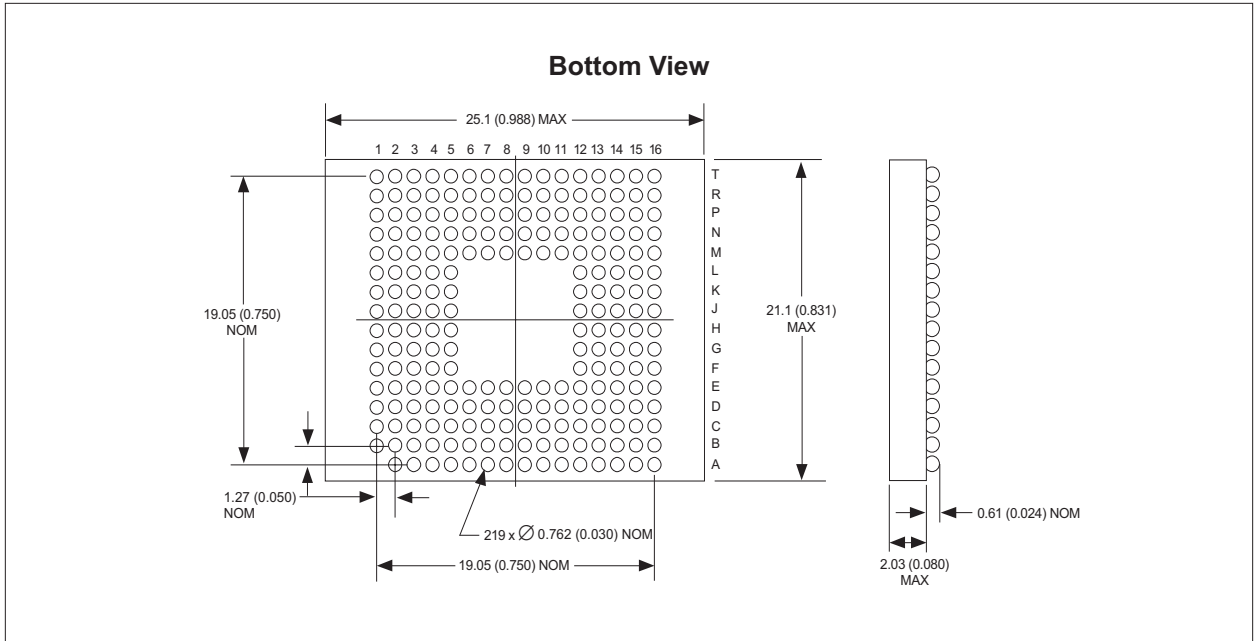
FIGURE C – PULL-DOWN CHARACTERISTICS

FIGURE D – PULL-UP CHARACTERISTICS

PACKAGE DIMENSION: 219 PLASITC BALL GRID ARRAY (PBGA)

Bottom View

ALL LINEAR DIMENSIONS ARE MILLIMETERS AND PARENTHETICALLY IN INCHES

ORDERING INFORMATION

W 3E 16M 64 S - XXX B X

WHITE ELECTRONIC DESIGNS CORP.

DDR SDRAM

CONFIGURATION, 16M x 64

2.5V Power Supply

FREQUENCY (MHz)
200 = 200Mbps
250 = 250Mbps
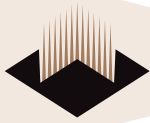**266 = 266Mbps**

PACKAGE:
B = 219 Plastic Ball Grid Array (PBGA)

DEVICE GRADE:
M = Military   -55°C to +125°C
I = Industrial   -40°C to +85°C
C = Commercial 0°C to +70°C

Document Title

16M x 64 DDR SDRAM Multi-Chip Package

Revision History

Rev #	History	Release Date	Status
Rev 0	Initial Release	December 2002	Advanced
Rev 1	Changes (Pg. 1, 15, 16) Package dimension changes 1.1 Ball diameter to 0.76 +/-0.08 1.2 Package size to 25x21mm Max 1.3 Package height to 2.65mm Max 1.4 Add solder ball coplanarity of 0.2mm	May 2003	Advanced
Rev 2	Changes (Pg. 1, 15, 16) 2.1 Change mechanical drawing to new style 2.2 Change status to preliminary	November 2003	Preliminary
Rev 3	Changes (Pg. 1, 10, 11, 12, 13, 15, 16) 3.1 Change status to Final 3.2 Correct typographical errors	September 2004	Final
Rev 4	Changes (Pg. 1, 11, 16) 4.1 Changes I_{OH}, I_{OL} to 12mA minimum 4.2 Update I_{CC} Specifications table values	February 2005	Final