

DATA SHEET

SAA7120; SAA7121 Digital Video Encoder (ConDENC)

Preliminary specification
File under Integrated Circuits, IC22

1997 Jan 06

Digital Video Encoder (ConDENC)**SAA7120; SAA7121****FEATURES**

- Monolithic CMOS 3.3 V (5 V) device
- Digital PAL/NTSC encoder
- System pixel frequency 13.5 MHz
- Accepts MPEG decoded data on 8-bit wide input port; input data format Cb-Y-Cr (CCIR 656), SAV and EAV
- Three DACs for Y, C and CVBS, two times oversampled with 10 bit resolution
- Real time control of subcarrier
- Cross colour reduction filter
- Closed captioning encoding and WST- and NABTS-Teletext encoding including sequencer and filter
- Line 23 wide screen signalling encoding
- Fast I²C-bus control port (400 kHz)
- Encoder can be master or slave
- Programmable horizontal and vertical input synchronization phase
- Programmable horizontal sync output phase
- Internal colour bar generator (CBG)
- 2 × 2 bytes in lines 20 (NTSC) for copy guard management system can be loaded via I²C-bus
- Down-mode of DACs
- Controlled rise/fall times of synchronization and blanking output signals



- Macrovision Pay-per-View copy protection system rev.7 and rev.6.1 as option.

This applies to SAA7120 only. The device is protected by USA patent numbers 4631603, 4577216 and 4819098 and other intellectual property rights. Use of the Macrovision anti-copy process in the device is licensed for non-commercial home use only. Reverse engineering or disassembly is prohibited. Please contact your nearest Philips Semiconductors sales office for more information.

- QFP44 package.

GENERAL DESCRIPTION

The SAA7120; SAA7121 encodes digital YUV video data to an NTSC or PAL CVBS or S-Video signal.

The circuit accepts CCIR compatible YUV data with 720 active pixels per line in 4 : 2 : 2 multiplexed formats, for example MPEG decoded data. It includes a sync/clock generator and on-chip DACs.

QUICK REFERENCE DATA

| SYMBOL | PARAMETER | MIN. | TYP. | MAX. | UNIT |
|---------------------|--|----------------|------|------|------|
| V _{DDA} | analog supply voltage | 3.1 | 3.3 | 3.5 | V |
| V _{DDD} | digital supply voltage | 3.0 | 3.3 | 3.6 | V |
| I _{DDA} | analog supply current | – | – | 62 | mA |
| I _{DDD} | digital supply current | – | – | 38 | mA |
| V _i | input signal voltage levels | TTL compatible | | | |
| V _{o(p-p)} | analog output signal voltages Y, C, and CVBS without load (peak-to-peak value) | 1.2 | 1.35 | 1.45 | V |
| R _L | load resistance | 75 | – | 300 | Ω |
| ILE | LF integral linearity error | – | – | ±3 | LSB |
| DLE | LF differential linearity error | – | – | ±1 | LSB |
| T _{amb} | operating ambient temperature | 0 | – | +70 | °C |

Digital Video Encoder (ConDENC)

SAA7120; SAA7121

ORDERING INFORMATION

| TYPE NUMBER | PACKAGE | | |
|---------------------|---------|--|----------|
| | NAME | DESCRIPTION | VERSION |
| SAA7120; SAA7121 | QFP44 | plastic quad flat package; 44 leads (lead length 2.35 mm); body 10 × 10 × 1.75 mm | SOT307-2 |

BLOCK DIAGRAM

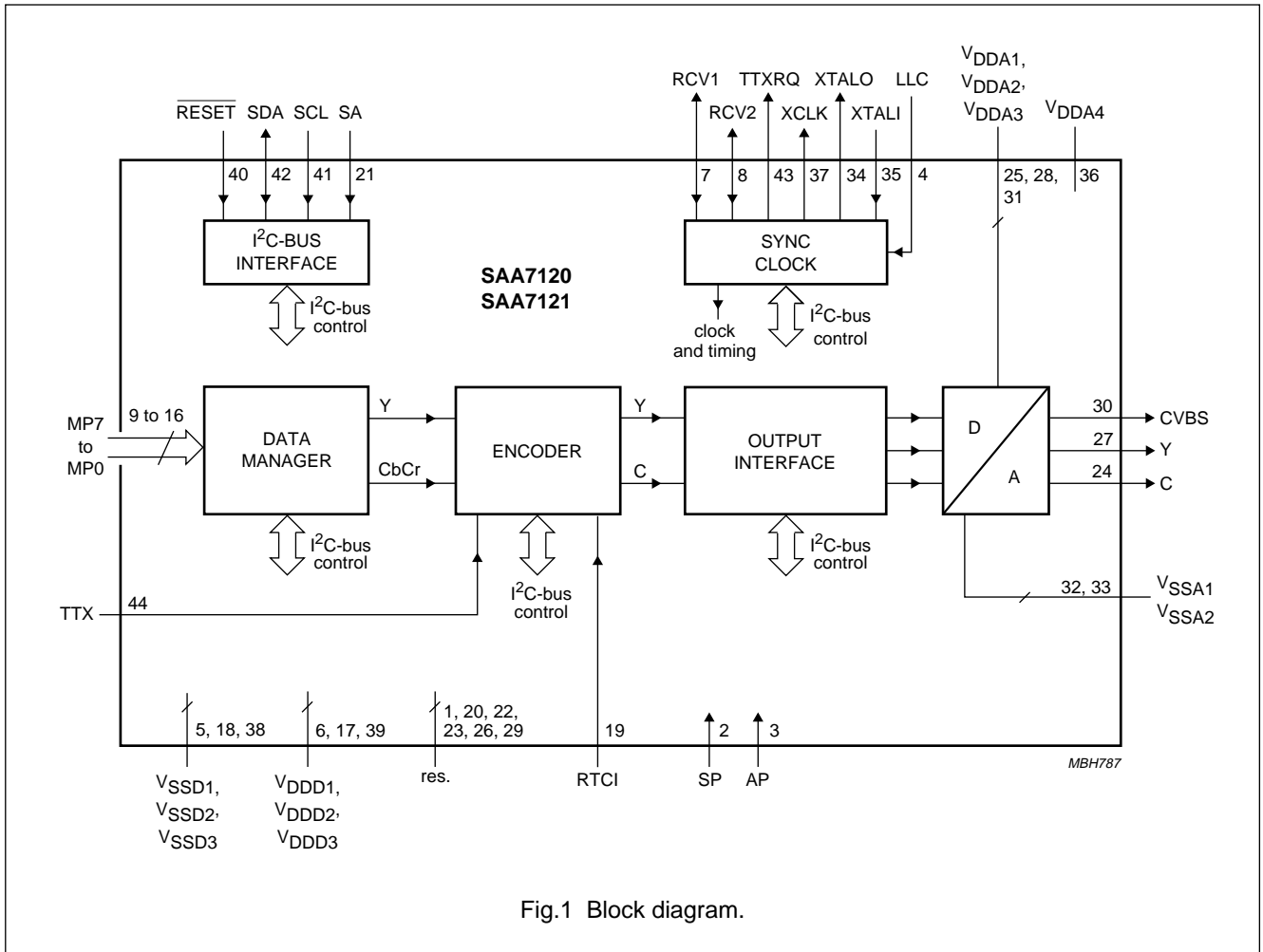


Fig.1 Block diagram.

Digital Video Encoder (ConDENC)

SAA7120; SAA7121

PINNING

| SYMBOL | PIN | I/O | DESCRIPTION |
|-------------------|-----|-----|--|
| res. | 1 | – | reserved |
| SP | 2 | I | test pin; connected to digital ground for normal operation |
| AP | 3 | I | test pin; connected to digital ground for normal operation |
| LLC | 4 | I | line-locked clock; this is the 27 MHz master clock for the encoder |
| V _{SSD1} | 5 | I | digital ground 1 |
| V _{DDD1} | 6 | I | digital supply voltage 1 |
| RCV1 | 7 | I/O | raster control 1 for video port; this pin receives/provides a VS/FS/FSEQ signal |
| RCV2 | 8 | I/O | raster control 2 for video port; this pin provides an HS pulse of programmable length or receives an HS pulse |
| MP7 | 9 | I | MPEG port; it is an input for "CCIR 656" style multiplexed Cb Y, Cr data |
| MP6 | 10 | I | |
| MP5 | 11 | I | |
| MP4 | 12 | I | |
| MP3 | 13 | I | |
| MP2 | 14 | I | |
| MP1 | 15 | I | |
| MP0 | 16 | I | |
| V _{DDD2} | 17 | I | digital supply voltage 2 |
| V _{SSD2} | 18 | I | digital ground 2 |
| RTCI | 19 | I | Real Time Control input; if the LLC clock is provided by an SAA7111 or SAA7151B, RTCI should be connected to pin RTCO of the decoder to improve the signal quality |
| res. | 20 | – | reserved |
| SA | 21 | I | the I ² C-bus slave address select input pin; LOW: slave address = 88H, HIGH = 8CH |
| res. | 22 | – | reserved |
| res. | 23 | – | reserved |
| C | 24 | O | analog output of the chrominance signal |
| V _{DDA1} | 25 | I | analog supply voltage 1 for the C DAC |
| res. | 26 | – | reserved |
| Y | 27 | O | analog output of VBS signal |
| V _{DDA2} | 28 | I | analog supply voltage 2 for the Y DAC |
| res. | 29 | – | reserved |
| CVBS | 30 | O | analog output of the CVBS signal |
| V _{DDA3} | 31 | I | analog supply voltage 3 for the CVBS DAC |
| V _{SSA1} | 32 | I | analog ground 1 for the DACs |
| V _{SSA2} | 33 | I | analog ground 2 for the oscillator and reference voltage |
| XTALO | 34 | O | crystal oscillator output (to crystal) |
| XTALI | 35 | I | crystal oscillator input (from crystal); if the oscillator is not used, this pin should be connected to ground |
| V _{DDA4} | 36 | I | analog supply voltage 4 for the oscillator and reference voltage |
| XCLK | 37 | O | clock output of the crystal oscillator |

Digital Video Encoder (ConDENC)

SAA7120; SAA7121

| SYMBOL | PIN | I/O | DESCRIPTION |
|-------------------|-----|-----|--|
| V _{SSD3} | 38 | I | digital ground 3 |
| V _{DDD3} | 39 | I | digital supply voltage 3 |
| RESET | 40 | I | reset input, active LOW; after reset is applied, all digital I/Os are in input mode; the I ² C-bus receiver waits for the START condition |
| SCL | 41 | I | I ² C-bus serial clock input |
| SDA | 42 | I/O | I ² C-bus serial data input/output |
| TTXRQ | 43 | O | teletext request output, indicating when bit stream is valid |
| TTX | 44 | I | teletext bit stream input |

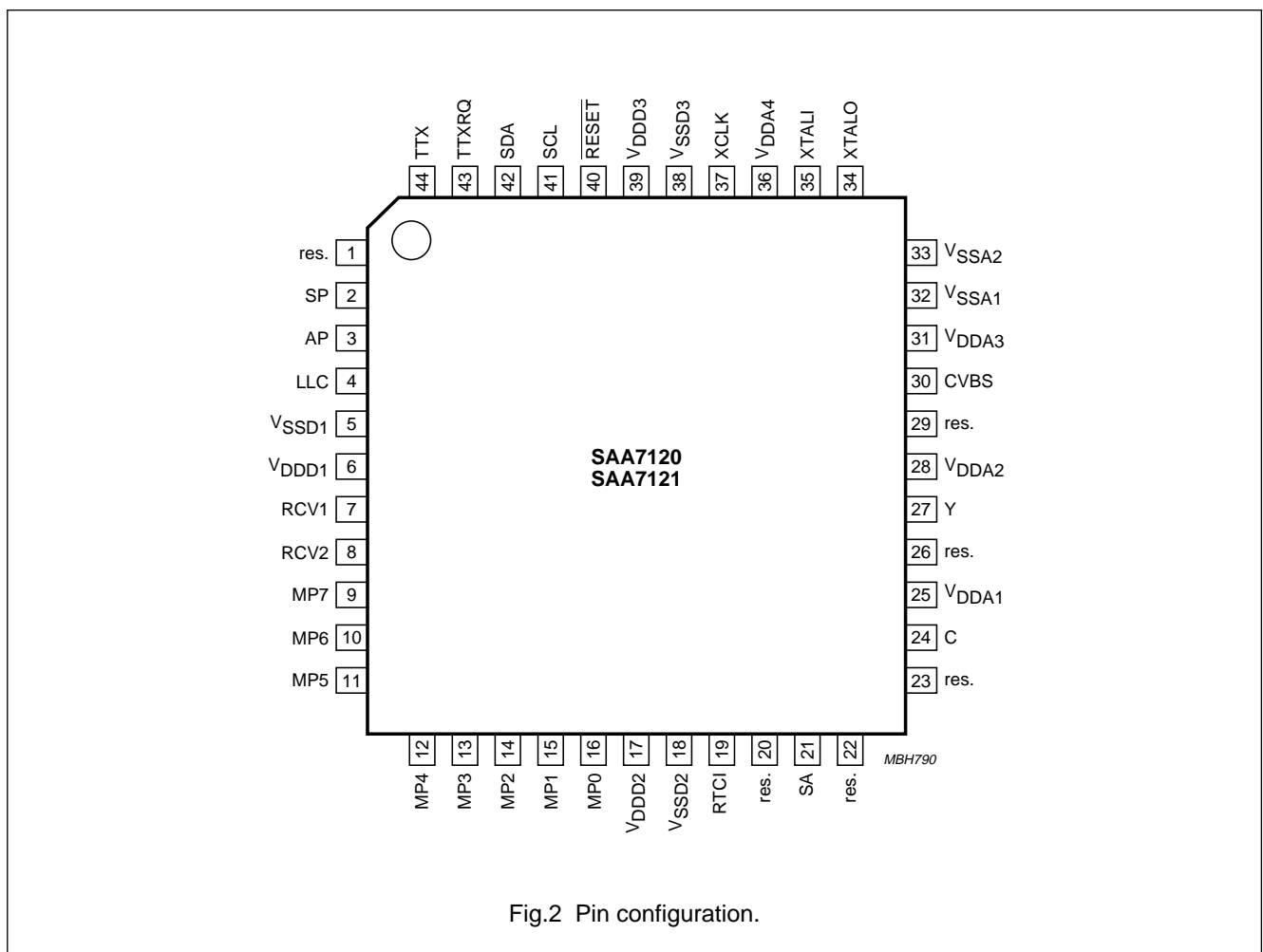


Fig.2 Pin configuration.

Digital Video Encoder (ConDENC)

SAA7120; SAA7121

FUNCTIONAL DESCRIPTION

The digital video encoder (ConDENC) encodes digital luminance and colour difference signals simultaneously into analog CVBS and S-Video signals. NTSC-M, PAL B/G, and sub-standards are supported.

Both interlaced and non-interlaced operation is possible for all standards.

The basic encoder function consists of subcarrier generation, colour modulation and the insertion of synchronization signals. Luminance and chrominance signals are filtered in accordance with the standard requirements of "RS-170-A" and "CCIR 624".

For ease of analog post-filtering the signals are oversampled twice with respect to the pixel clock prior to digital-to-analog conversion.

The filter characteristics are shown in Figs 3 and 4. The DACs for Y, C, and CVBS have 10-bit resolution.

The 8-bit multiplexed Cb-Y-Cr formats are "CCIR 656" (D1 format) compatible, but the SAV and EAV codes can

be decoded optionally when the device is to operate in slave mode.

It is also possible to connect a Philips Digital Video Decoder (SAA7111 or SAA7151B) to the ConDENC. Via pin RTCI, connected to RTCO of a decoder, information concerning the actual subcarrier, PAL-ID and (if used in conjunction with the SAA7111) the subcarrier phase can be inserted.

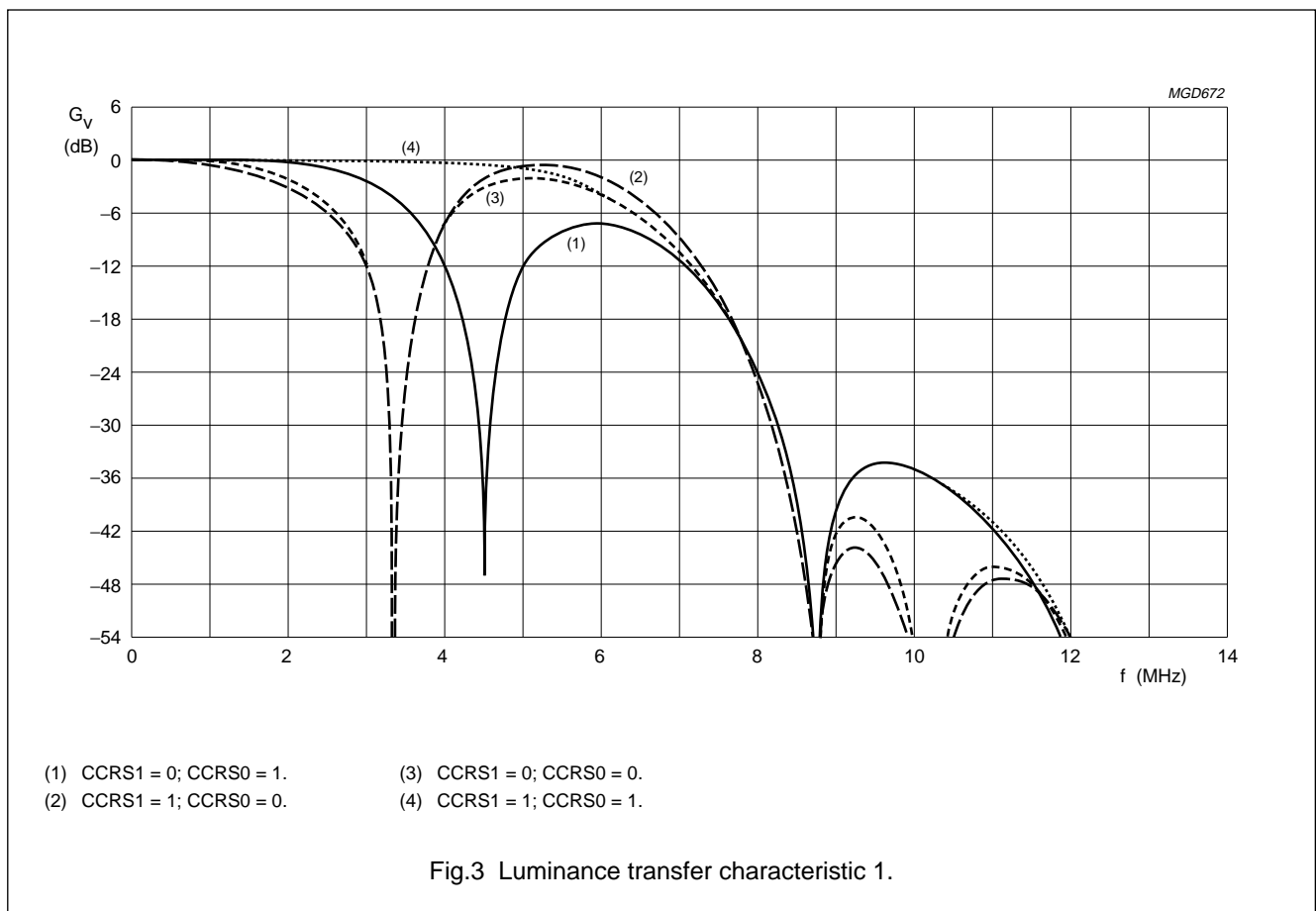
The ConDENC synthesizes all necessary internal signals, colour subcarrier frequency and synchronization signals.

Wide screen signalling data can be loaded via the I²C-bus. It is inserted into line 23 for 50 Hz field rate standards.

The IC contains closed caption and extended data services encoding (line 21), and supports anti-taping signal generation in accordance with Macrovision.

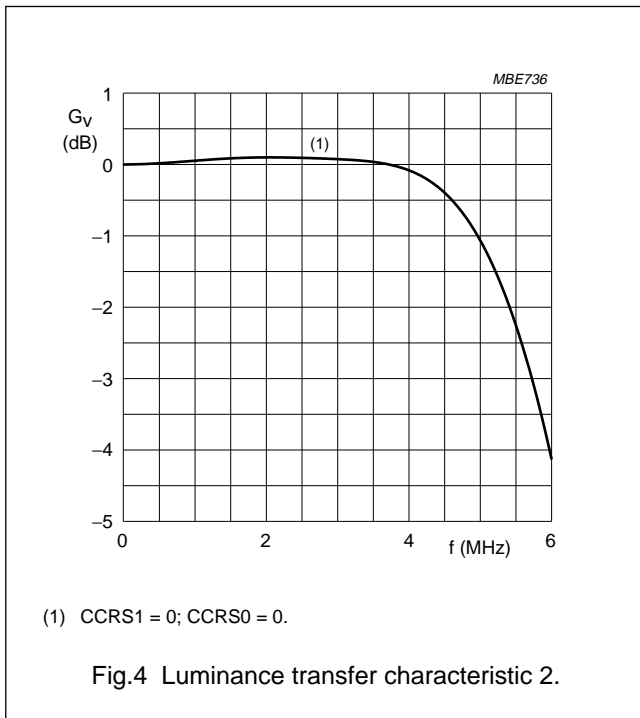
Possibilities are provided for setting video parameters:

- Black and blanking level control
- Colour subcarrier frequency
- Variable burst amplitude.



Digital Video Encoder (ConDENC)

SAA7120; SAA7121



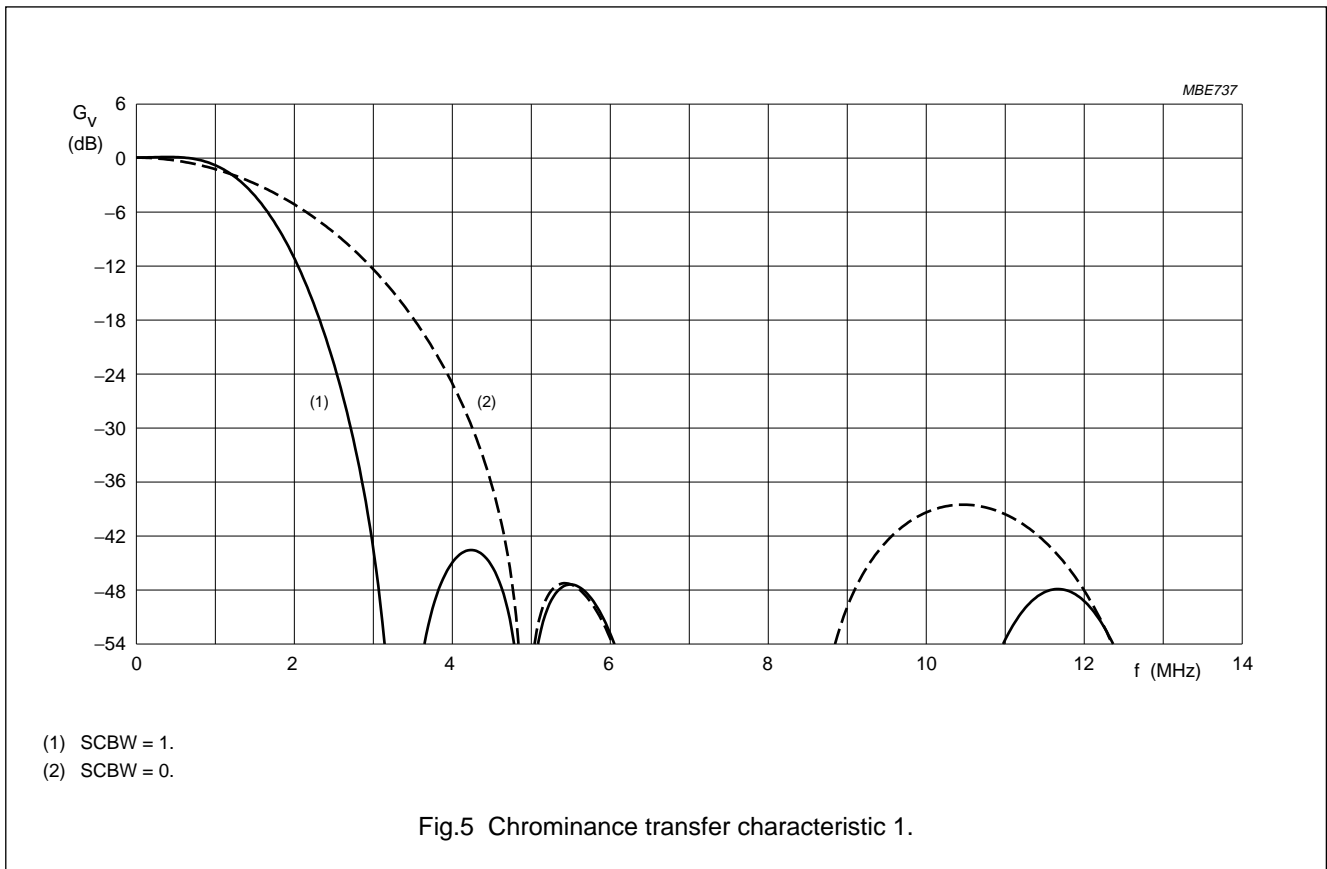
During reset ($\overline{\text{RESET}} = \text{LOW}$) and after reset is released, all digital I/O stages are set to input mode. A reset forces the I²C-bus interface to abort a running bus transfer and sets register 3A to 03H, register 61 to 06H, registers 6BH and 6EH to 00H and bit TTX60 to 0. All other control registers are not influenced by a reset.

Encoder

VIDEO PATH

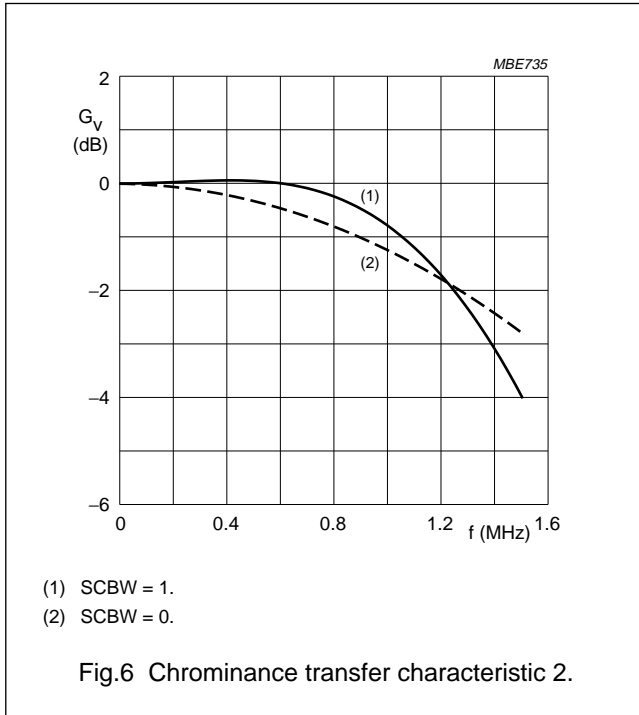
The encoder generates out of Y, U and V baseband signals luminance and colour subcarrier output signals, suitable for use as CVBS or separate Y and C signals.

Luminance is modified in gain and in offset (the latter programmable in a certain range to enable different black level set-ups). A fixed synchronization level in accordance with standard composite synchronization schemes is inserted. The inserted blanking level is programmable to allow for manipulations with Macrovision anti-taping. Additional insertion of AGC super-white pulses, programmable in height, is supported.



Digital Video Encoder (ConDENC)

SAA7120; SAA7121



In order to enable easy analog post-filtering, luminance is interpolated from 13.5 MHz data rate to 27 MHz data rate, providing luminance in 10-bit resolution. This filter is also used to define smoothed transients for synchronization pulses and blanking period. For transfer characteristic of the luminance interpolation filter see Figs 3 and 4.

Chrominance is modified in gain (programmable separately for U and V), standard dependent burst is inserted, before baseband colour signals are interpolated from 6.75 MHz data rate to 27 MHz data rate. One of the interpolation stages can be bypassed, thus providing a higher colour bandwidth, which can be made use of for Y and C output. For transfer characteristics of the chrominance interpolation filter see Figs 5 and 6.

The amplitude, beginning and ending of inserted burst is programmable in a certain range, suitable for standard signals and for special effects. Behind the succeeding quadrature modulator, colour in 10-bit resolution is provided on subcarrier.

The numeric ratio between Y and C outputs is in accordance with set standards.

TELETEXT INSERTION AND ENCODING

Pin TTX receives a WST- or NABTS-Teletext bitstream sampled at the LLC clock. At each rising edge of output

signal TTXRQ a single teletext bit has to be provided after a programmable delay at input pin.

Phase variant interpolation is achieved on this bitstream in the internal teletext encoder, providing sufficient small phase jitter on the output text lines.

TTXRQ provides a fully programmable request signal to the teletext source, indicating the insertion period of bitstream at lines selectable independently for both fields. The internal insertion window for text is set to 360 (PAL-WST), 296 (NTSC-WST) or 288 (NABTS) teletext bits including clock run-in bits. For protocol and timing see Fig.7.

CLOSED CAPTION ENCODER

Using this circuit, data in accordance with the specification of closed caption or extended data service, delivered by the control interface, can be encoded (Line 21). Two dedicated pairs of bytes (two bytes per field), each pair preceded by run-in clocks and framing code, are possible.

The actual line number where data is to be encoded in, can be modified in a certain range.

Data clock frequency is in accordance with definition for NTSC-M standard 32 times horizontal line frequency.

Data LOW at the output of the DACs corresponds to 0 IRE, data HIGH at the output of the DACs corresponds to approximately 50 IRE.

It is also possible to encode Closed Caption Data for 50 Hz field frequencies at 32 times horizontal line frequency.

ANTI-TAPING (SAA7120 ONLY)

For more information contact your nearest Philips Semiconductors sales office.

Data manager

In the data manager, real time arbitration on the data stream to be encoded is performed.

A pre-defined colour look-up table located in this block can be read out in a pre-defined sequence (8 steps per active video line), achieving a colour bar test pattern generator without the need for an external data source. The colour bar function is under software control only.

Output interface/DACs

In the output interface encoded Y and C signals are converted from digital to analog in 10-bit resolution.

Digital Video Encoder (ConDENC)

SAA7120; SAA7121

Y and C signals are also combined to a 10-bit CVBS signal.

The CVBS output occurs with the same processing delay as the Y and C outputs. Absolute amplitude at the input of the DAC for CVBS is reduced by $15/16$ with respect to Y and C DACs to make maximum use of conversion ranges.

Outputs of the DACs can be set together in two groups via software control to minimum output voltage for either purpose.

Synchronization

Synchronization of the ConDENC is able to operate in two modes; slave mode and master mode.

In the slave mode, the circuit accepts synchronization pulses at the bidirectional RCV1 port. The timing and trigger behaviour related to RCV1 can be influenced by programming the polarity and the on-chip delay of RCV1. Active slope of RCV1 defines the vertical phase and optionally the odd/even and colour frame phase to be initialized, it can be also used to set the horizontal phase.

If the horizontal phase is not to be influenced by RCV1, a horizontal synchronization pulse needs to be supplied at the pin RCV2. Timing and trigger behaviour can also be influenced by RCV2.

If there are missing pulses at RCV1 and/or RCV2, the time base of ConDENC runs free, thus an arbitrary number of synchronization slopes may be absent, but no additional pulses (with the incorrect phase) must occur.

If the vertical and horizontal phase is derived from RCV1, RCV2 can be used for horizontal or composite blanking input or output.

Alternatively, the device can be triggered by auxiliary codes in a "CCIR 656" data stream at the MP port.

In the master mode, the time base of the circuit continuously runs free. On the RCV1 port, the device can output:

- A Vertical Synchronisation signal (VS) with 3 or 2.5 lines duration, or
- An ODD/EVEN signal which is LOW in odd fields, or
- A field sequence signal (FSEQ) which is HIGH in the first of 4 or 8 fields respectively.

On the RCV2 port, the device can provide a horizontal synchronization pulse with programmable start and stop phase; this pulse can be inhibited in the vertical blanking period to build up, for example, a composite blanking signal.

The polarity of both RCV1 and RCV2 is selectable by software control.

The length of a field and the start and end of its active part can be programmed. The active part of a field always starts at the beginning of a line.

Teletext timing

The teletext timing is shown in Fig.7. t_{FD} is the time needed to interpolate input data TTX and inserting it into the CVBS and Y output signal, such that it appears at $t_{TTX} = 10.2 \mu\text{s}$ (PAL) or $t_{TTX} = 10.5 \mu\text{s}$ (NTSC) after the leading edge of the horizontal synchronization pulse.

Time t_{pD} is the pipeline delay time introduced by the source that is gated by TTXRQ in order to deliver TTX data. This delay is programmable by register TTXHD. For every active HIGH-state at output pin TTXRQ, a new teletext bit must be provided by the source.

Since the beginning of the pulses representing the TTXRQ signal and the delay between the rising edge of TTXRQ and valid teletext input data are fully programmable (TTXHS and TTXHD), the TTX data is always inserted at the correct position after the leading edge of outgoing horizontal synchronization pulse.

Time t_{TTXWin} is the internally used insertion window for TTX data; it has a constant length that allows insertion of 360 teletext bits at a text data rate of 6.9375 Mb/s (PAL), 296 teletext bits at a text data rate of 5.7272 Mb/s (World Standard TTX) or 288 teletext bits at a text data rate of 5.7272 Mb/s (NABTS). The insertion window is not opened if the control bit TTXEN is logic 0.

Using appropriate programming, all suitable lines of the odd field (TTXOVS and TTXOVE) plus all suitable lines of the even field (TTXEVS and TTXEVE) can be used for teletext insertion.

Digital Video Encoder (ConDENC)

SAA7120; SAA7121

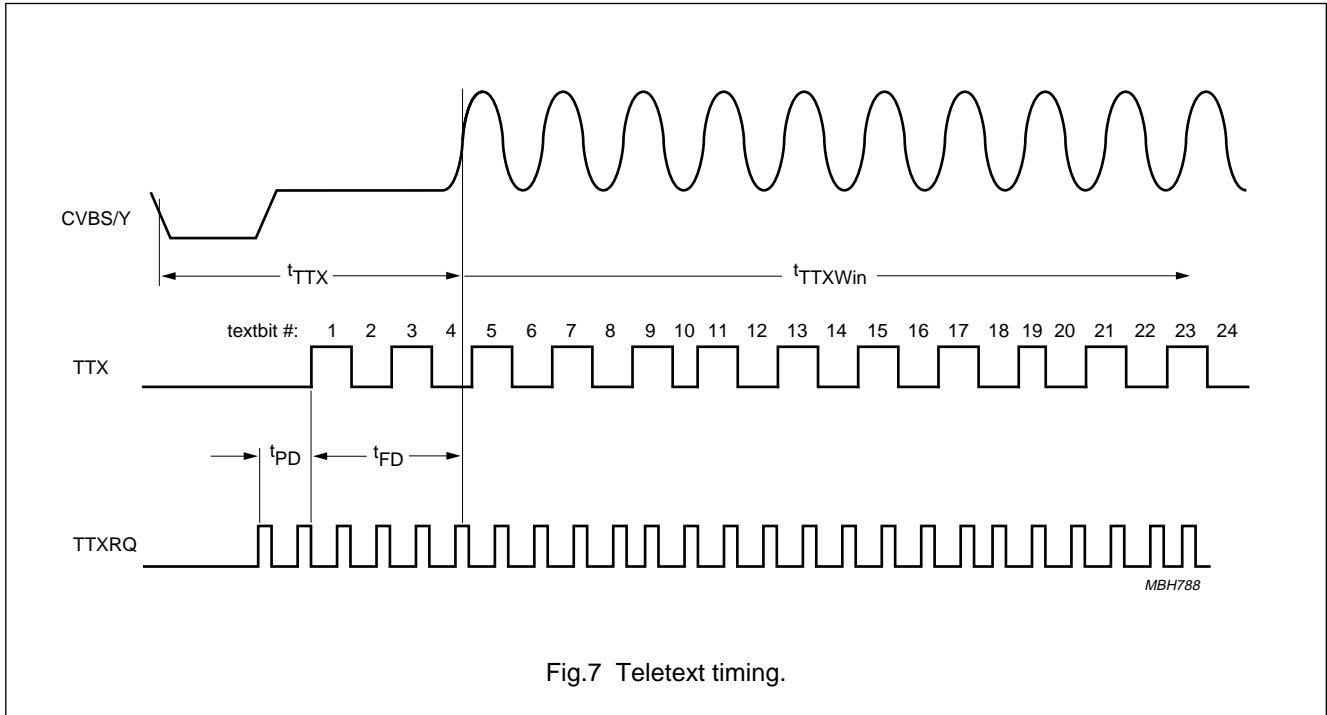


Fig.7 Teletext timing.

Analog output voltages

The analog output voltages are dependent on the open-loop voltage of the operational amplifiers for full-scale conversion (typical value 1.35 V), the internal series resistor (typical value 2 Ω), the external series resistor and the external load impedance.

The digital output signals in front of the DACs under nominal conditions occupy different conversion ranges, as indicated in Table 1 for a 100/100 colour bar signal.

Values for the external series resistors result in a 75 Ω load.

Input levels and formats

The ConDENC expects digital Y, Cb, Cr data with levels (digital codes) in accordance with "CCIR 601" (see Tables 2 and 3).

For C and CVBS outputs, deviating amplitudes of the colour difference signals can be compensated by independent gain control setting, while gain for luminance is set to predefined values, distinguishable for 7.5 IRE set-up or without set-up.

Reference levels are measured with a colour bar, 100% white, 100% amplitude and 100% saturation.

Table 1 Digital output signals conversion range

| CONVERSION RANGE (peak-to-peak) (digits) | |
|--|---------------------------|
| CVBS, SYNC TIP-TO-PEAK CARRIER | Y (VBS) SYNC TIP-TO-WHITE |
| 1016 | 881 |

Table 2 "CCIR 601" signal component levels

| COLOUR | SIGNALS | | |
|---------|---------|-----|-----|
| | Y | Cb | Cr |
| White | 235 | 128 | 128 |
| Yellow | 210 | 16 | 146 |
| Cyan | 170 | 166 | 16 |
| Green | 145 | 54 | 34 |
| Magenta | 106 | 202 | 222 |
| Red | 81 | 90 | 240 |
| Blue | 41 | 240 | 110 |
| Black | 16 | 128 | 128 |

Digital Video Encoder (ConDENC)

SAA7120; SAA7121

I²C-bus interface

The I²C-bus interface is a standard slave transceiver, supporting 7-bit slave addresses and 400 kbits/s guaranteed transfer rate. It uses 8-bit subaddressing with an auto-increment function. All registers are write only, except one readable status byte.

Two I²C-bus slave addresses are present:

88H: LOW at pin SA

8CH: HIGH at pin SA.

Tables 5 and 4 summarize the format of the I²C-bus addressing. For more information on how to use the I²C-bus see *"The I²C-bus and how to use it"*, order no. 9398 393 40011. Tables 7 to 42 contain the programming information for the subaddresses. Table 6 summarises this information.

Table 3 8-bit multiplexed format (similar to "CCIR 601")

| | BITS | | | | | | | |
|------------------------|-----------------|----------------|-----------------|----------------|-----------------|----------------|-----------------|----------------|
| | 0 | 1 | 2 | 2 | 4 | 5 | 6 | 7 |
| Sample | Cb ₀ | Y ₀ | Cr ₀ | Y ₁ | Cb ₂ | Y ₂ | Cr ₂ | Y ₃ |
| Luminance pixel number | 0 | | 1 | | 2 | | 3 | |
| Colour pixel number | 0 | | | | 2 | | | |

Table 4 I²C-bus address format; see Table 5

| | | | | | | | | | | |
|---|---------------|-----|------------|-----|--------|-----|-------|--------|-----|---|
| S | SLAVE ADDRESS | ACK | SUBADDRESS | ACK | DATA 0 | ACK | ----- | DATA n | ACK | P |
|---|---------------|-----|------------|-----|--------|-----|-------|--------|-----|---|

Table 5 Explanation of Table 4

| PART | DESCRIPTION |
|---------------------------|---|
| S | START condition |
| Slave address | 1 0 0 0 1 0 0 x or 1 0 0 0 1 1 0 x ⁽¹⁾ |
| ACK | acknowledge, generated by the slave |
| Subaddress ⁽²⁾ | subaddress byte |
| DATA | data byte |
| ----- | continued data bytes and ACKs |
| P | STOP condition |

Notes

- x is the read/write control bit; write:
x = logic 0;
read: x = logic 1, no subaddressing with read.
- If more than 1 byte DATA is transmitted, then auto-increment of the subaddress is performed.

Digital Video Encoder (ConDENC)

SAA7120; SAA7121

Table 6 Slave receiver (slave address 88H or 8CH)

| REGISTER FUNCTION | SUB ADDRESS | DATA BITS ⁽¹⁾ | | | | | | | | | | |
|---|-------------|--------------------------|---------|--------|--------|--------|--------|--------|--------|---|---|---|
| | | D7 | D6 | D5 | D4 | D3 | D2 | D1 | D0 | | | |
| Null | 00 | 0 | 0 | 0 | 0 | 0 | 0 | 0 | 0 | 0 | 0 | 0 |
| Null | 25 | 0 | 0 | 0 | 0 | 0 | 0 | 0 | 0 | 0 | 0 | 0 |
| Wide screen signal | 26 | WSS7 | WSS6 | WSS5 | WSS4 | WSS3 | WSS2 | WSS1 | WSS0 | | | |
| Wide screen signal | 27 | WSSON | 0 | WSS13 | WSS12 | WSS11 | WSS10 | WSS9 | WSS8 | | | |
| Real time control, Burst start | 28 | DECCOL | DEC FIS | BS5 | BS4 | BS3 | BS2 | BS1 | BS0 | | | |
| Burst end | 29 | 0 | 0 | BE5 | BE4 | BE3 | BE2 | BE1 | BE0 | | | |
| Copy guard odd 0 | 2A | CGO07 | CGO06 | CGO05 | CGO04 | CGO03 | CGO02 | CGO01 | CGO00 | | | |
| Copy guard odd 1 | 2B | CGO17 | CGO16 | CGO15 | CGO14 | CGO13 | CGO12 | CGO11 | CGO10 | | | |
| Copy guard even 0 | 2C | CGE07 | CGE06 | CGE05 | CGE04 | CGE03 | CGE02 | CGE01 | CGE00 | | | |
| Copy guard even 1 | 2D | CGE17 | CGE16 | CGE15 | CGE14 | CGE13 | CGE12 | CGE11 | CGE10 | | | |
| Copy guard enable | 2E | CGEN1 | CGENO | 0 | 0 | 0 | 0 | 0 | 0 | | | |
| Null | 2F | 0 | 0 | 0 | 0 | 0 | 0 | 0 | 0 | | | |
| Null | 39 | 0 | 0 | 0 | 0 | 0 | 0 | 0 | 0 | | | |
| Input port control | 3A | CBENB | 0 | 0 | SYMP | 0 | 0 | Y2C | UV2C | | | |
| Chrominance phase | 5A | CHPS7 | CHPS6 | CHPS5 | CHPS4 | CHPS3 | CHPS2 | CHPS1 | CHPS0 | | | |
| Gain U | 5B | GAINU7 | GAINU6 | GAINU5 | GAINU4 | GAINU3 | GAINU2 | GAINU1 | GAINU0 | | | |
| Gain V | 5C | GAINV7 | GAINV6 | GAINV5 | GAINV4 | GAINV3 | GAINV2 | GAINV1 | GAINV0 | | | |
| Gain U MSB, Real time control, Black level | 5D | GAINU8 | DECOE | BLCKL5 | BLCKL4 | BLCKL3 | BLCKL2 | BLCKL1 | BLCKL0 | | | |
| Gain V MSB, Real time control, Blanking level | 5E | GAINV8 | DECPH | BLNNL5 | BLNNL4 | BLNNL3 | BLNNL2 | BLNNL1 | BLNNL0 | | | |
| CCR, Blanking level VBI | 5F | CCRS1 | CCRS0 | BLNVB5 | BLNVB4 | BLNVB3 | BLNVB2 | BLNVB1 | BLNVB0 | | | |
| Null | 60 | 0 | 0 | 0 | 0 | 0 | 0 | 0 | 0 | | | |
| Standard control | 61 | 0 | DOWN | INPI | YGS | 0 | SCBW | PAL | FISE | | | |
| RTC enable, Burst amplitude | 62 | RTCE | BSTA6 | BSTA5 | BSTA4 | BSTA3 | BSTA2 | BSTA1 | BSTA0 | | | |
| Subcarrier 0 | 63 | FSC07 | FSC06 | FSC05 | FSC04 | FSC03 | FSC02 | FSC01 | FSC00 | | | |
| Subcarrier 1 | 64 | FSC15 | FSC14 | FSC13 | FSC12 | FSC11 | FSC10 | FSC09 | FSC08 | | | |
| Subcarrier 2 | 65 | FSC23 | FSC22 | FSC21 | FSC20 | FSC19 | FSC18 | FSC17 | FSC16 | | | |

Digital Video Encoder (ConDENC)

SAA7120; SAA7121

| REGISTER FUNCTION | SUB ADDRESS | DATA BITS ⁽¹⁾ | | | | | | | | | |
|---------------------------------|-------------|--------------------------|---------|---------|---------|---------|---------|---------|---------|--|--|
| | | D7 | D6 | D5 | D4 | D3 | D2 | D1 | D0 | | |
| Subcarrier 3 | 66 | FSC31 | FSC30 | FSC29 | FSC28 | FSC27 | FSC26 | FSC25 | FSC24 | | |
| Line 21 odd 0 | 67 | L21O07 | L21O06 | L21O05 | L21O04 | L21O03 | L21O02 | L21O01 | L21O00 | | |
| Line 21 odd 1 | 68 | L21O17 | L21O16 | L21O15 | L21O14 | L21O13 | L21O12 | L21O11 | L21O10 | | |
| Line 21 even 0 | 69 | L21E07 | L21E06 | L21E05 | L21E04 | L21E03 | L21E02 | L21E01 | L21E00 | | |
| Line 21 even 1 | 6A | L21E17 | L21E16 | L21E15 | L21E14 | L21E13 | L21E12 | L21E11 | L21E10 | | |
| RCV port control | 6B | SRCV11 | SRCV10 | TRCV2 | ORCV1 | PRCV1 | CBLF | ORCV2 | PRCV2 | | |
| Trigger control | 6C | HTRIG7 | HTRIG6 | HTRIG5 | HTRIG4 | HTRIG3 | HTRIG2 | HTRIG1 | HTRIG0 | | |
| Trigger control | 6D | HTRIG10 | HTRIG9 | HTRIG8 | VTRIG4 | VTRIG3 | VTRIG2 | VTRIG1 | VTRIG0 | | |
| Multi control | 6E | SBLBN | 0 | PHRES1 | PHRES0 | 0 | 0 | FLC1 | FLC0 | | |
| Closed caption, Teletext enable | 6F | CCEN1 | CCEN0 | TTXEN | SCCLN4 | SCCLN3 | SCCLN2 | SCCLN1 | SCCLN0 | | |
| RCV2 output start | 70 | RCV2S7 | RCV2S6 | RCV2S5 | RCV2S4 | RCV2S3 | RCV2S2 | RCV2S1 | RCV2S0 | | |
| RCV2 output end | 71 | RCV2E7 | RCV2E6 | RCV2E5 | RCV2E4 | RCV2E3 | RCV2E2 | RCV2E1 | RCV2E0 | | |
| MSBs RCV2 output | 72 | 0 | RCV2E10 | RCV2E9 | RCV2E8 | 0 | RCV2S10 | RCV2S9 | RCV2S8 | | |
| TTX request H start | 73 | TTXHS7 | TTXHS6 | TTXHS5 | TTXHS4 | TTXHS3 | TTXHS2 | TTXHS1 | TTXHS0 | | |
| TTX request H delay | 74 | TTXHD7 | TTXHD6 | TTXHD5 | TTXHD4 | TTXHD3 | TTXHD2 | TTXHD1 | TTXHD0 | | |
| V-Sync shift | 75 | 0 | 0 | 0 | 0 | 0 | VS_S2 | VS_S1 | VS_S0 | | |
| TTX odd request V S | 76 | TTXOVS7 | TTXOVS6 | TTXOVS5 | TTXOVS4 | TTXOVS3 | TTXOVS2 | TTXOVS1 | TTXOVS0 | | |
| TTX odd request V E | 77 | TTXOVE7 | TTXOVE6 | TTXOVE5 | TTXOVE4 | TTXOVE3 | TTXOVE2 | TTXOVE1 | TTXOVE0 | | |
| TTX even request V S | 78 | TTXEVS7 | TTXEVS6 | TTXEVS5 | TTXEVS4 | TTXEVS3 | TTXEVS2 | TTXEVS1 | TTXEVS0 | | |
| TTX even request V E | 79 | TTXEVE7 | TTXEVE6 | TTXEVE5 | TTXEVE4 | TTXEVE3 | TTXEVE2 | TTXEVE1 | TTXEVE0 | | |
| First active line | 7A | FAL7 | FAL6 | FAL5 | FAL4 | FAL3 | FAL2 | FAL1 | FAL0 | | |
| Last active line | 7B | LAL7 | LAL6 | LAL5 | LAL4 | LAL3 | LAL2 | LAL1 | LAL0 | | |
| MSB vertical | 7C | TTX60 | LAL8 | 0 | FAL8 | TTXEVE8 | TTXOVE8 | TTXEVS8 | TTXOVS8 | | |
| Null | 7D | 0 | 0 | 0 | 0 | 0 | 0 | 0 | 0 | | |
| Disable TTX line | 7E | LINE12 | LINE11 | LINE10 | LINE9 | LINE8 | LINE7 | LINE6 | LINE5 | | |
| Disable TTX line | 7F | LINE20 | LINE19 | LINE18 | LINE17 | LINE16 | LINE15 | LINE14 | LINE13 | | |

Note

1. All bits labelled '0' are reserved. They **must** be programmed with logic 0.

Digital Video Encoder (ConDENC)

SAA7120; SAA7121

Slave Receiver

Table 7 Subaddress 26 and 27

| DATA BYTE | LOGIC LEVEL | DESCRIPTION |
|-----------|-------------|---|
| WSS | – | wide screen signalling bits: 13 to 11 = reserved 10 to 8 = subtitles 7 to 4 = enhanced services 3 to 0 = aspect ratio |
| WSSON | 0 | wide screen signalling output is disabled |
| | 1 | wide screen signalling output is enabled |

Table 8 Subaddress 28 and 29

| DATA BYTE | LOGIC LEVEL | DESCRIPTION | REMARKS |
|-----------|-------------|--|---|
| BS | – | starting point of burst in clock cycles | PAL : BS = 33 (21H) NTSC : BS = 25 (19H) |
| BE | – | ending point of burst in clock cycles | PAL : BS = 29 (1DH) NTSC : BS = 29 (1DH) |
| DECCOL | 0 | disable colour detection bit of RTCI input | |
| | 1 | enable colour detection bit of RTCI input | bit RTCE must be set to 1 (see Fig.10) |
| DEC FIS | 0 | field sequence as FISE in subaddress 61 | |
| | 1 | field sequence as FISE bit in RTCI input | bit RTCE must be set to 1 (see Fig.10) |

Table 9 Subaddress 2A to 2D

| DATA BYTE | DESCRIPTION | REMARKS |
|-----------|--|--|
| CGO0 | first byte of Copy guard data, odd field | LSBs of the respective bytes are encoded immediately after run-in and framing code, the MSBs of the respective bytes have to carry the parity bit, in accordance with the definition of Line 20 encoding format. |
| CGO1 | second byte of Copy guard data, odd field | |
| CGE0 | first byte of Copy guard data, even field | |
| CGE1 | second byte of Copy guard data, even field | |

Table 10 Subaddress 2E

| DATA BYTE | | DESCRIPTION |
|-----------|-------|------------------------------------|
| CCEN1 | CCEN0 | |
| 0 | 0 | copy guard encoding off |
| 0 | 1 | enables encoding in field 1 (odd) |
| 1 | 0 | enables encoding in field 2 (even) |
| 1 | 1 | enables encoding in both fields |

Digital Video Encoder (ConDENC)

SAA7120; SAA7121

Table 11 Subaddress 3A

| DATA BYTE | LOGIC LEVEL | DESCRIPTION |
|-----------|-------------|---|
| UV2C | 0 | Cb, Cr data are two's complement |
| | 1 | Cb, Cr data are straight binary; default after reset |
| Y2C | 0 | Y data is two's complement |
| | 1 | Y data is straight binary; default after reset |
| SYMP | 0 | horizontal and vertical trigger is taken from RCV2 and RCV1 respectively; default after reset |
| | 1 | horizontal and vertical trigger is decoded out of "CCIR 656" compatible data at MP port |
| CBENB | 0 | data from input ports is encoded; default after reset |
| | 1 | colour bar with fixed colours is encoded |

Table 12 Subaddress 5A

| DATA BYTE | DESCRIPTION | VALUE | RESULT |
|-----------|---|-------|-------------------------------------|
| CHPS | phase of encoded colour subcarrier (including burst) relative to horizontal sync; can be adjusted in steps of 360/256 degrees | 3FH | PAL-B/G and data from input ports |
| | | 69H | PAL-B/G and data from look-up table |
| | | 67H | NTSC-M and data from input ports |
| | | 89H | NTSC-M and data from look-up table |

Remark: in subaddresses 5B, 5C, 5D, 5E and 62 all IRE values are rounded up.

Table 13 Subaddress 5B and 5D

| DATA BYTE | DESCRIPTION | CONDITIONS | REMARKS |
|-----------|--|---|--|
| GAINU | variable gain for Cb signal; input representation accordance with "CCIR 601" | white-to-black = 92.5 IRE GAINU = 0 GAINU = 118 (76H) | GAINU = $-2.17 \times \text{nominal}$ to $+2.16 \times \text{nominal}$ output subcarrier of U contribution = 0 output subcarrier of U contribution = nominal |
| | | white-to-black = 100 IRE GAINU = 0 GAINU = 125 (7DH) | GAINU = $-2.05 \times \text{nominal}$ to $+2.04 \times \text{nominal}$ output subcarrier of U contribution = 0 output subcarrier of U contribution = nominal |

Table 14 Subaddress 5C and 5E

| DATA BYTE | DESCRIPTION | CONDITIONS | REMARKS |
|-----------|--|---|--|
| GAINV | variable gain for Cr signal; input representation accordance with "CCIR 601" | white-to-black = 92.5 IRE GAINV = 0 GAINV = 165 (A5H) | GAINV = $-1.55 \times \text{nominal}$ to $+1.55 \times \text{nominal}$ output subcarrier of V contribution = 0 output subcarrier of V contribution = nominal |
| | | white-to-black = 100 IRE GAINV = 0 GAINV = 175 (AFH) | GAINV = $-1.46 \times \text{nominal}$ to $+1.46 \times \text{nominal}$ output subcarrier of V contribution = 0 output subcarrier of V contribution = nominal |

Digital Video Encoder (ConDENC)

SAA7120; SAA7121

Table 15 Subaddress 5D

| DATA BYTE | DESCRIPTION | CONDITIONS | REMARKS |
|-----------|---|---|---|
| BLCKL | variable black level; input representation accordance with "CCIR 601" | white-to-sync = 140 IRE ⁽¹⁾ BLCKL = 0 BLCKL = 63 (3FH) | recommended value: BLCKL = 42 (2AH) output black level = 34 IRE output black level = 54 IRE |
| | | white-to-sync = 143 IRE ⁽²⁾ BLCKL = 0 BLCKL = 63 (3FH) | recommended value: BLCKL = 35 (23H) output black level = 32 IRE output black level = 52 IRE |
| DECOE | real time control | logic 0 | disable odd/even field control bit from RTCI |
| | | logic 1 | enable odd/even field control bit from RTCI (see Fig.10) |

Notes

- Output black level/IRE = $BLCKL \times 2/6.29 + 34.0$
- Output black level/IRE = $BLCKL \times 2/6.18 + 31.7$

Table 16 Subaddress 5E

| DATA BYTE | DESCRIPTION | CONDITIONS | REMARKS |
|-----------|-------------------------|---|---|
| BLNNL | variable blanking level | white-to-sync = 140 IRE ⁽¹⁾ BLNNL = 0 BLNNL = 63 (3FH) | recommended value: BLNNL = 46 (2EH) output blanking level = 25 IRE output blanking level = 45 IRE |
| | | white-to-sync = 143 IRE ⁽²⁾ BLNNL = 0 BLNNL = 63 (3FH) | recommended value: BLNNL = 53 (35H) output blanking level = 26 IRE output blanking level = 46 IRE |
| DECPH | real time control | logic 0 | disable subcarrier phase reset bit from RTCI |
| | | logic 1 | enable subcarrier phase reset bit from RTCI (see Fig.10) |

Notes

- Output black level/IRE = $BLNNL \times 2/6.29 + 25.4$
- Output black level/IRE = $BLNNL \times 2/6.18 + 25.9$

Table 17 Subaddress 5F

| DATA BYTE | DESCRIPTION |
|-----------|--|
| BLNVB | variable blanking level during vertical blanking interval is typically identical to value of BLNNL |
| CCRS | select cross colour reduction filter in luminance; see Table 18 |

Table 18 Logic levels and function of CCRS

| CCRS1 | CCRS0 | DESCRIPTION |
|-------|-------|---|
| 0 | 0 | no cross colour reduction; for overall transfer characteristic of luminance see Fig.3 |
| 0 | 1 | cross colour reduction #1 active; for overall transfer characteristic see Fig.3 |
| 1 | 0 | cross colour reduction #2 active; for overall transfer characteristic see Fig.3 |
| 1 | 1 | cross colour reduction #3 active; for overall transfer characteristic see Fig.3 |

Digital Video Encoder (ConDENC)

SAA7120; SAA7121

Table 19 Subaddress 61

| DATA BYTE | LOGIC LEVEL | DESCRIPTION |
|-----------|-------------|---|
| FISE | 0 | 864 total pixel clocks per line; default after reset |
| | 1 | 858 total pixel clocks per line |
| PAL | 0 | NTSC encoding (non-alternating V component) |
| | 1 | PAL encoding (alternating V component); default after reset |
| SCBW | 0 | enlarged bandwidth for chrominance encoding (for overall transfer characteristic of chrominance in baseband representation see Figs 3 and 4) |
| | 1 | standard bandwidth for chrominance encoding (for overall transfer characteristic of chrominance in baseband representation see Figs 3 and 4); default after reset |
| YGS | 0 | luminance gain for white – black 100 IRE; default after reset |
| | 1 | luminance gain for white – black 92.5 IRE including 7.5 IRE set-up of black |
| INPI | 0 | PAL switch phase is nominal; default after reset |
| | 1 | PAL switch phase is inverted compared to nominal |
| DOWN | 0 | DACs for CVBS, Y and C in normal operational mode; default after reset |
| | 1 | DACs for CVBS, Y and C forced to lowest output voltage |

Table 20 Subaddress 62H

| DATA BYTE | LOGIC LEVEL | DESCRIPTION |
|-----------|-------------|---|
| RTCE | 0 | no real time control of generated subcarrier frequency |
| | 1 | real time control of generated subcarrier frequency through SAA7151B or SAA7111 (timing see Fig.10) |

Table 21 Subaddress 62H

| DATA BYTE | DESCRIPTION | CONDITIONS | REMARKS |
|-----------|---|--|---------------------------------------|
| BSTA | amplitude of colour burst; input representation in accordance with "CCIR 601" | white-to-black = 92.5 IRE; burst = 40 IRE; NTSC encoding BSTA = 0 to $2.02 \times$ nominal | recommended value: BSTA = 63 (3FH) |
| | | white-to-black = 92.5 IRE; burst = 40 IRE; PAL encoding BSTA = 0 to $2.82 \times$ nominal | recommended value: BSTA = 45 (2DH) |
| | | white-to-black = 100 IRE; burst = 43 IRE; NTSC encoding BSTA = 0 to $1.90 \times$ nominal | recommended value: BSTA = 67 (43H) |
| | | white-to-black = 100 IRE; burst = 43 IRE; PAL encoding BSTA = 0 to $3.02 \times$ nominal | recommended value: BSTA = 47 (2FH) |

Digital Video Encoder (ConDENC)

SAA7120; SAA7121

Table 22 Subaddress 63 to 66 (four bytes to program subcarrier frequency)

| DATA BYTE | DESCRIPTION | CONDITIONS | REMARKS |
|--------------|--|---|---|
| FSC0 to FSC3 | f_{fsc} = subcarrier frequency (in multiples of line frequency); f_{llc} = clock frequency (in multiples of line frequency) | $FSC = \left(\frac{f_{fsc}}{f_{llc}} \times 2^{32} \right),$ rounded up; see note 1 | FSC3 = most significant byte FSC0 = least significant byte |

Note

1. Examples:

- a) NTSC-M: $f_{fsc} = 227.5$, $f_{llc} = 1716 \rightarrow FSC = 569408543$ (21F07C1FH).
b) PAL-B/G: $f_{fsc} = 283.7516$, $f_{llc} = 1728 \rightarrow FSC = 705268427$ (2A098ACBH).

Table 23 Subaddress 67 to 6A

| DATA BYTE | DESCRIPTION | REMARKS |
|-----------|---|--|
| L21O0 | first byte of captioning data, odd field | LSBs of the respective bytes are encoded immediately after run-in and framing code, the MSBs of the respective bytes have to carry the parity bit, in accordance with the definition of Line 21 encoding format. |
| L21O1 | second byte of captioning data, odd field | |
| L21E0 | first byte of extended data, even field | |
| L21E1 | second byte of extended data, even field | |

Table 24 Subaddress 6B

| DATA BYTE | LOGIC LEVEL | DESCRIPTION |
|-----------|-------------|---|
| PRCV2 | 0 | polarity of RCV2 as output is active HIGH, rising edge is taken when input, respectively; default after reset |
| | 1 | polarity of RCV2 as output is active LOW, falling edge is taken when input, respectively |
| ORCV2 | 0 | pin RCV2 is switched to input; default after reset |
| | 1 | pin RCV2 is switched to output |
| CBLF | 0 | if ORCV2 = HIGH, pin RCV2 provides an HREF signal (Horizontal Reference pulse that is defined by RCV2S and RCV2E, also during vertical blanking Interval); default after reset if ORCV2 = LOW and bit SYMP = LOW, signal input to RCV2 is used for horizontal synchronization only (if TRCV2 = 1); default after reset |
| | 1 | if ORCV2 = HIGH, pin RCV2 provides a 'Composite-Blanking-Not' signal, for example a reference pulse that is defined by RCV2S and RCV2E, excluding Vertical Blanking Interval, which is defined by FAL and LAL if ORCV2 = LOW and bit SYMP = LOW, signal input to RCV2 is used for horizontal synchronization (if TRCV2 = 1) and as an internal blanking signal |
| PRCV1 | 0 | polarity of RCV1 as output is active HIGH, rising edge is taken when input; default after reset |
| | 1 | polarity of RCV1 as output is active LOW, falling edge is taken when input |
| ORCV1 | 0 | pin RCV1 is switched to input; default after reset |
| | 1 | pin RCV1 is switched to output |
| TRCV2 | 0 | horizontal synchronization is taken from RCV1 port (at bit SYMP = LOW) or from decoded frame sync of "CCIR 656" input (at bit SYMP = HIGH); default after reset |
| | 1 | horizontal synchronization is taken from RCV2 port (at bit SYMP = LOW) |
| SRCV1 | – | defines signal type on pin RCV1; see Table 25 |

Digital Video Encoder (ConDENC)

SAA7120; SAA7121

Table 25 Logic levels and function of SRCV1

| DATA BYTE | | AS OUTPUT | AS INPUT | FUNCTION |
|-----------|--------|----------------|----------------|--|
| SRCV11 | SRCV10 | | | |
| 0 | 0 | VS | VS | vertical sync each field; default after reset |
| 0 | 1 | FS | FS | frame sync (odd/even) |
| 1 | 0 | FSEQ | FSEQ | field sequence, vertical sync every fourth field (PAL = 0) or eighth field (PAL = 1) |
| 1 | 1 | not applicable | not applicable | – |

Table 26 Subaddress 6C and 6D

| DATA BYTE | DESCRIPTION |
|-----------|--|
| HTRIG | sets the horizontal trigger phase related to signal on RCV1 or RCV2 input values above 1715 (FISE = 1) or [1727 (FISE = 0)] are not allowed increasing HTRIG decreases delays of all internally generated timing signals reference mark: analog output horizontal sync (leading slope) coincides with active edge of RCV used for triggering at HTRIG = 398H [398H] |

Table 27 Subaddress 6D

| DATA BYTE | DESCRIPTION |
|-----------|---|
| VTRIG | sets the vertical trigger phase related to signal on RCV1 input increasing VTRIG decreases delays of all internally generated timing signals, measured in half lines variation range of VTRIG = 0 to 31 (1FH) |

Table 28 Subaddress 6E

| DATA BYTE | LOGIC LEVEL | DESCRIPTION |
|-----------|-------------|---|
| SBLBN | 0 | vertical blanking is defined by programming of FAL and LAL; default after reset |
| | 1 | vertical blanking is forced in accordance with "CCIR 624" (50 Hz) or RS170A (60 Hz) |
| PHRES | – | selects the phase reset mode of the colour subcarrier generator; see Table 29 |
| FLC | – | field length control; see Table 30 |

Table 29 Logic levels and function of PHRES

| DATA BYTE | | DESCRIPTION |
|-----------|--------|--|
| PHRES1 | PHRES0 | |
| 0 | 0 | no reset or reset via RTCI from SAA7111 if bit RTCE = 1; default after reset |
| 0 | 1 | reset every two lines |
| 1 | 0 | reset every eight fields |
| 1 | 1 | reset every four fields |

Digital Video Encoder (ConDENC)

SAA7120; SAA7121

Table 30 Logic levels and function of FLC

| DATA BYTE | | DESCRIPTION |
|-----------|------|--|
| FLC1 | FLC0 | |
| 0 | 0 | interlaced 312.5 lines/field at 50 Hz, 262.5 lines/field at 60 Hz; default after reset |
| 0 | 1 | non-interlaced 312 lines/field at 50 Hz, 262 lines/field at 60 Hz |
| 1 | 0 | non-interlaced 313 lines/field at 50 Hz, 263 lines/field at 60 Hz |
| 1 | 1 | non-interlaced 313 lines/field at 50 Hz, 263 lines/field at 60 Hz |

Table 31 Subaddress 6F

| DATA BYTE | LOGIC LEVEL | DESCRIPTION |
|-----------|-------------|--|
| CCEN | – | enables individual Line 21 encoding; see Table 32 |
| TTXEN | 0 | disables teletext insertion |
| | 1 | enables teletext insertion |
| SCCLN | – | selects the actual line, where closed caption or extended data are encoded line = (SCCLN + 4) for M-systems line = (SCCLN + 1) for other systems |

Table 32 Logic levels and function of CCEN

| DATA BYTE | | DESCRIPTION |
|-----------|-------|------------------------------------|
| CCEN1 | CCEN0 | |
| 0 | 0 | Line 21 encoding off |
| 0 | 1 | enables encoding in field 1 (odd) |
| 1 | 0 | enables encoding in field 2 (even) |
| 1 | 1 | enables encoding in both fields |

Table 33 Subaddress 70 to 72

| DATA BYTE | DESCRIPTION |
|-----------|---|
| RCV2S | start of output signal on pin RCV2 values above 1715 (FISE = 1) or [1727 (FISE = 0)] are not allowed first active pixel at analog outputs (corresponding input pixel coinciding with RCV2) at RCV2S = 11AH [0FDH] |
| RCV2E | end of output signal on pin RCV2 values above 1715 (FISE = 1) or [1727 (FISE = 0)] are not allowed last active pixel at analog outputs (corresponding input pixel coinciding with RCV2) at RCV2E = 694H [687H] |

Digital Video Encoder (ConDENC)

SAA7120; SAA7121

Table 34 Subaddress 73 and 74

| DATA BYTE | DESCRIPTION |
|-----------|--|
| TTXHS | start of signal on pin TTXRQ see Fig.7 |
| TTXHD | indicates the delay in clock cycles between rising edge of TTXRQ output and valid data on pin TTX minimum value has to be TTXHD = 2 |

Table 35 Subaddress 75

| DATA BYTE | DESCRIPTION |
|-----------|--|
| VS_S | Vertical Sync. shift between RCV1 and RCV2 (switched to output) in master mode it is possible to shift H-sync (RCV2; CBLF = 0) against V-sync (RCV1; SRCV1 = 00) standard value: VS_S = 3 |

Table 36 Subaddress 76, 77 and 7C

| DATA BYTE | DESCRIPTION |
|-----------|--|
| TTXOVS | first line of occurrence of signal on pin TTXRQ in odd field line = (TTXOVS + 4) for M-systems line = (TTXOVS + 1) for other systems |
| TTXOVE | last line of occurrence of signal on pin TTXRQ in odd field line = (TTXOVE + 3) for M-systems line = TTXOVE for other systems |

Table 37 Subaddress 78, 79 and 7C

| DATA BYTE | DESCRIPTION |
|-----------|---|
| TTXEVS | first line of occurrence of signal on pin TTXRQ in even field line = (TTXEVS + 4) for M-systems line = (TTXEVS + 1) for other systems |
| TTXEVE | last line of occurrence of signal on pin TTXRQ in even field line = (TTXEVE + 3) for M-systems line = TTXEVE for other systems |

Table 38 Subaddress 7C

| DATA BYTE | LOGIC LEVEL | DESCRIPTION |
|-----------|-------------|--|
| TTX60 | 0 | enables NABTS (FISE = 1) or European TTX (FISE = 0); default after reset |
| | 1 | enables World Standard Teletext 60 Hz (FISE = 1) |

Table 39 Subaddress 7A to 7C

| DATA BYTE | DESCRIPTION |
|-----------|--|
| FAL | first active line = FAL + 4 for M-systems, = FAL + 1 for other systems, measured in lines FAL = 0 coincides with the first field synchronization pulse |
| LAL | last active line = LAL + 3 for M-systems, = LAL for other system, measured in lines LAL = 0 coincides with the first field synchronization pulse |

Digital Video Encoder (ConDENC)

SAA7120; SAA7121

Table 40 Subaddress 7E and 7F

| DATA BYTE | DESCRIPTION |
|-----------|--|
| LINE | individual lines in both fields (PAL counting) can be disabled for insertion of teletext by the respective bits, disabled line = LINE _{xx} (50 Hz field rate) this bit mask is effective only, if the lines are enabled by TTXOVS/TTXOVE and TTXEVS/TTXEVE |

Slave Transmitter**Table 41** Slave transmitter (slave address 89H or 8DH)

| REGISTER FUNCTION | SUBADDRESS | DATA BYTE | | | | | | | |
|-------------------|------------|-----------|------|------|-------|-------|----|------|-----|
| | | D7 | D6 | D5 | D4 | D3 | D2 | D1 | D0 |
| Status byte | – | VER2 | VER1 | VER0 | CCRDO | CCRDE | 0 | FSEQ | O_E |

Table 42 No subaddress

| DATA BYTE | LOGIC LEVEL | DESCRIPTION |
|-----------|-------------|---|
| VER | – | Version identification of the device. It will be changed with all versions of the device that have different programming models. Current version is 000 binary. |
| CCRDO | 1 | Closed caption bytes of the odd field have been encoded. |
| | 0 | The bit is reset after information has been written to the subaddresses 67 and 68. It is set immediately after the data has been encoded. |
| CCRDE | 1 | Closed caption bytes of the even field have been encoded. |
| | 0 | The bit is reset after information has been written to the subaddresses 69 and 6A. It is set immediately after the data has been encoded. |
| FSEQ | 1 | During first field of a sequence (repetition rate: NTSC = 4 fields, PAL = 8 fields). |
| | 0 | Not first field of a sequence. |
| O_E | 1 | During even field. |
| | 0 | During odd field. |

Digital Video Encoder (ConDENC)

SAA7120; SAA7121

CHARACTERISTICS

$V_{DD} = 3.0$ to 3.6 V; $T_{amb} = 0$ to $+70$ °C; unless otherwise specified.

| SYMBOL | PARAMETER | CONDITIONS | MIN. | MAX. | UNIT |
|--|--|----------------------------|---------------|-----------------|------|
| Supply | | | | | |
| V_{DDA} | analog supply voltage | | 3.1 | 3.5 | V |
| V_{DDD} | digital supply voltage | | 3.0 | 3.6 | V |
| I_{DDA} | analog supply current | note 1 | – | 62 | mA |
| I_{DDD} | digital supply current | note 1 | – | 38 | mA |
| Inputs | | | | | |
| V_{IL} | LOW level input voltage (except SDA, SCL, AP, SP and XTALI) | | –0.5 | +0.8 | V |
| V_{IH} | HIGH level input voltage (except, SDA, SCL, AP, SP and XTALI) | | 2.0 | $V_{DDD} + 0.3$ | V |
| I_{LI} | input leakage current | | – | 1 | μA |
| C_i | input capacitance | clocks | – | 10 | pF |
| | | data | – | 8 | pF |
| | | I/Os at high impedance | – | 8 | pF |
| Outputs | | | | | |
| V_{OL} | LOW level output voltage (except SDA and XTALO) | $I_{OL} = 4$ mA | – | 0.4 | V |
| V_{OH} | HIGH level output voltage (except, SDA, and XTALO) | $I_{OH} = 4$ mA | $V_{DDD} - 4$ | – | V |
| I²C-bus; SDA and SCL | | | | | |
| V_{IL} | LOW level input voltage | | –0.5 | $V_{DDD} + 0.3$ | V |
| V_{IH} | HIGH level input voltage | | 2.3 | $V_{DDD} + 0.3$ | V |
| I_i | input current | $V_i = \text{LOW or HIGH}$ | –10 | +10 | μA |
| V_{OL} | LOW level output voltage (SDA) | $I_{OL} = 3$ mA | – | 0.4 | V |
| I_o | output current | during acknowledge | 3 | – | mA |
| Clock timing (LLC) | | | | | |
| T_{LLC} | cycle time | note 2 | 34 | 41 | ns |
| δ | duty factor t_{HIGH}/t_{LLC} | note 3 | 40 | 60 | % |
| t_r | rise time | note 2 | – | 5 | ns |
| t_f | fall time | note 2 | – | 6 | ns |
| Input timing | | | | | |
| $t_{SU;DAT}$ | input data set-up time (any pin except SCL, SDA, RESET, AP and SP) | | 6 | – | ns |
| $t_{HD;DAT}$ | input data hold time (any pin except SCL, SDA, RESET, AP and SP) | | 3 | – | ns |

Digital Video Encoder (ConDENC)

SAA7120; SAA7121

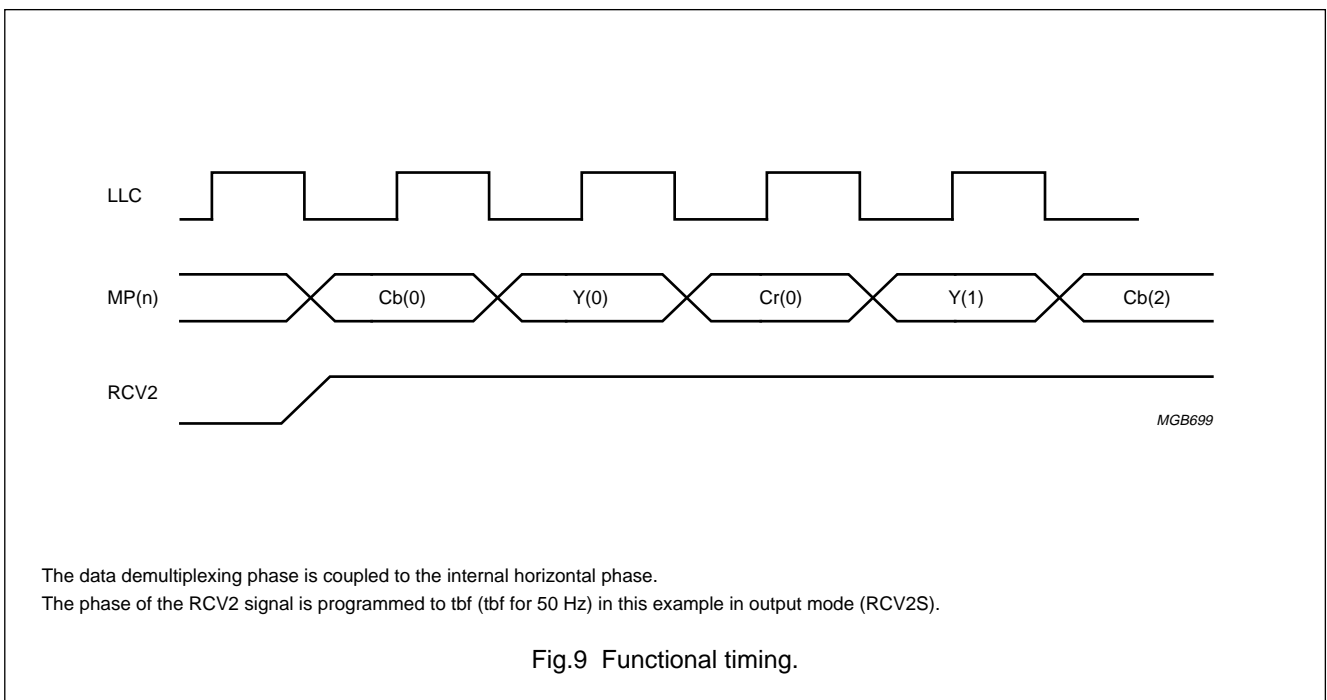
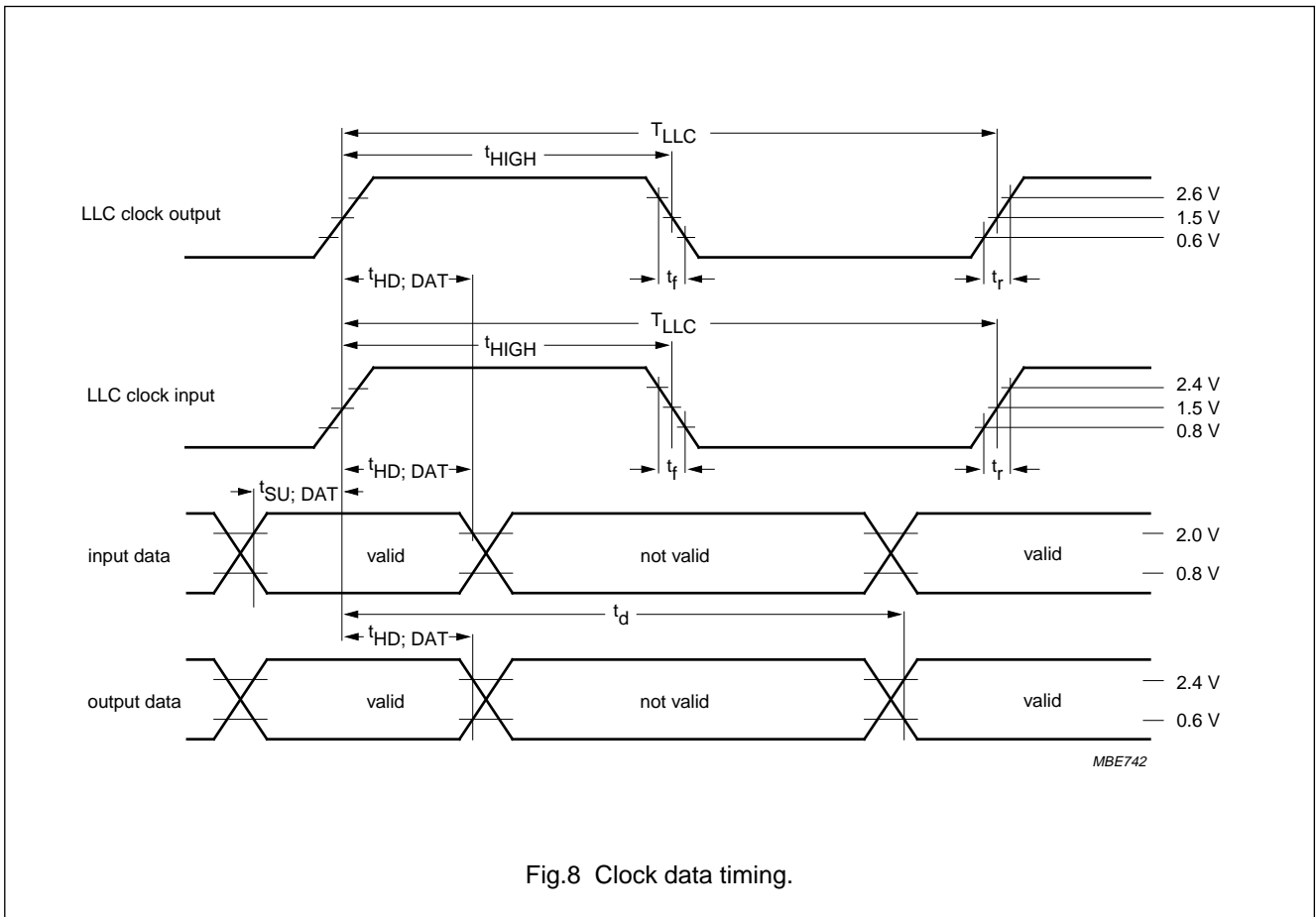
| SYMBOL | PARAMETER | CONDITIONS | MIN. | MAX. | UNIT |
|--|--|--------------------------|----------------------|----------------------|----------|
| Crystal oscillator | | | | | |
| f_n | nominal frequency (usually 27 MHz) | 3 rd harmonic | – | 30 | MHz |
| $\Delta f/f_n$ | permissible deviation of nominal frequency | note 4 | -50×10^{-6} | $+50 \times 10^{-6}$ | |
| CRYSTAL SPECIFICATION | | | | | |
| T_{amb} | operating ambient temperature | | 0 | 70 | °C |
| C_L | load capacitance | | 8 | – | pF |
| R_S | series resistance | | – | 80 | Ω |
| C_1 | motional capacitance (typical) | | 1.5 – 20% | 1.5 + 20% | fF |
| C_0 | parallel capacitance (typical) | | 3.5 – 20% | 3.5 + 20% | pF |
| Data and reference signal output timing | | | | | |
| C_L | output load capacitance | | 7.5 | 40 | pF |
| t_h | output hold time | | 4 | – | ns |
| t_d | output delay time | | – | 25 | ns |
| C, Y and CVBS outputs | | | | | |
| $V_{o(p-p)}$ | output signal voltage (peak-to-peak value) | note 5 | 1.20 | 1.45 | V |
| R_{int} | internal serial resistance | | 1 | 3 | Ω |
| R_L | output load resistance | | 75 | 300 | Ω |
| B_{-3dB} | output signal bandwidth of DACs | | 10 | – | MHz |
| ILE | LF integral linearity error of DACs | | – | ± 3 | LSB |
| DLE | LF differential linearity error of DACs | | – | ± 1 | LSB |

Notes

1. At maximum supply voltage with highly active input signals.
2. The data is for both input and output direction.
3. With LLC in input mode. In output mode, with a crystal connected to XTALO/XTALI duty factor is typically 50%.
4. If an internal oscillator is used, crystal deviation of nominal frequency is directly proportional to the deviation of subcarrier frequency and line/field frequency.
5. For full digital range, without load, $V_{DDA} = 3.3$ V. The typical voltage swing is 1.35 V, the typical minimum output voltage (digital zero at DAC) is 0.2 V.

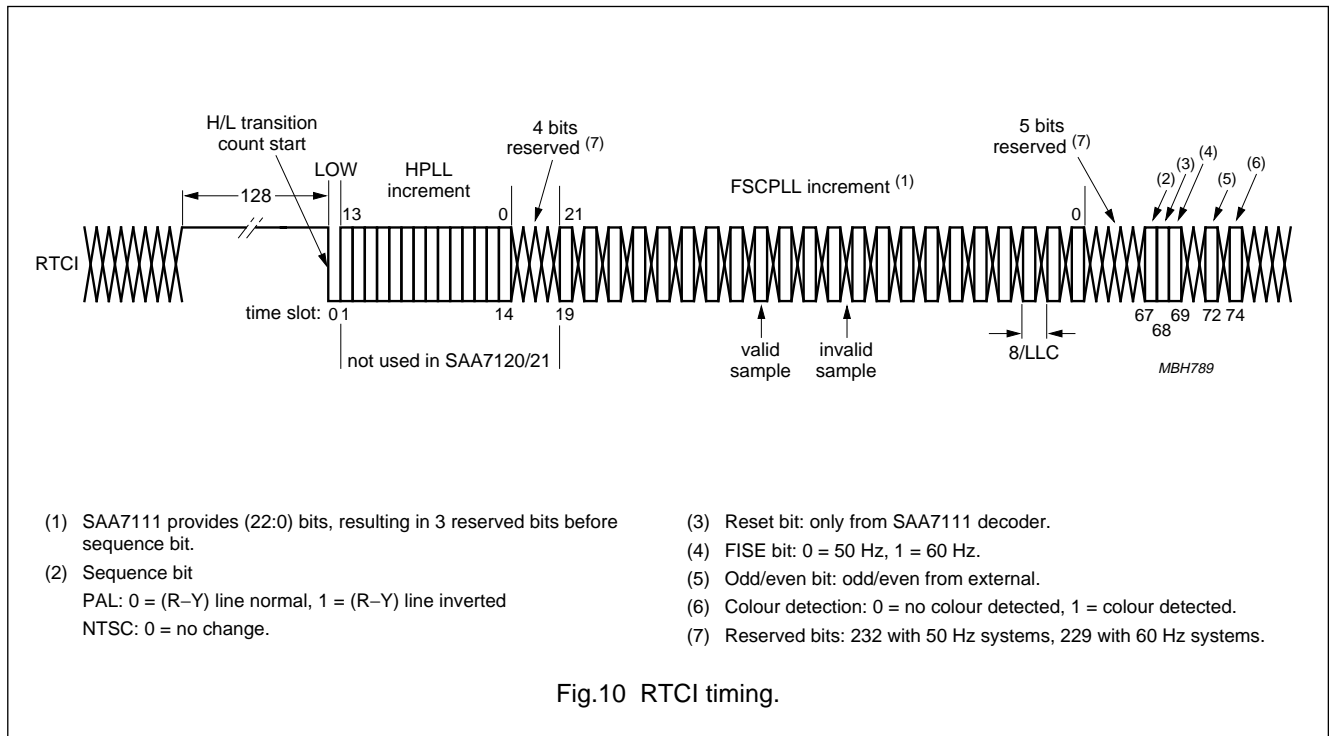
Digital Video Encoder (ConDENC)

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Explanation of RTCI data bits

1. The ConDENC generates the subcarrier frequency out of the FSCPLL increment if enabled (see item 6.).
2. The PAL bit indicates the line with inverted R - Y component of colour difference signal.
3. If the reset bit is enabled (RTCE = 1; DECPH = 1; PHRES = 00), the phase of the subcarrier is reset in each line whenever the reset bit of RTCI input is set to 1.
4. If the FISE bit is enabled (RTCE = 1; DECFIS = 1), the ConDENC takes this bit instead of the FISE bit in subaddress 61H.
5. If the odd/even bit is enabled (RTCE = 1; DECOE = 1), the ConDENC ignores its internally generated odd/even flag and takes the odd/even bit from RTCI input.

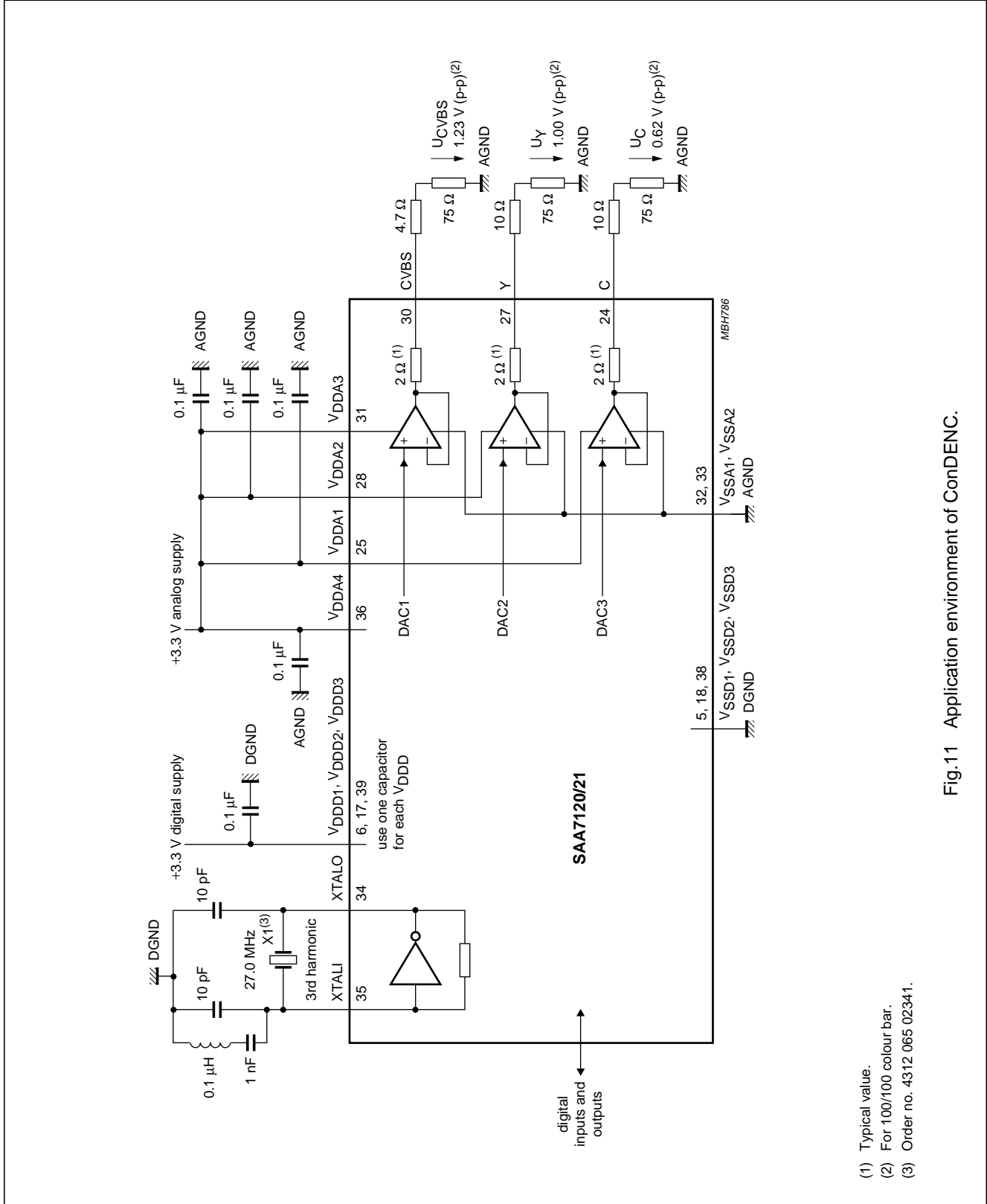
6. If the colour detection bit is enabled (RTCE = 1; DECCOL = 1) and no colour was detected (colour detection bit = 0), the subcarrier frequency is generated by the ConDENC. In the other case (colour detection bit = 1) the subcarrier frequency is evaluated out of FSCPLL increment.

If the colour detection bit is disabled (RTCE = 1; DECCOL = 0), the subcarrier frequency is evaluated out of FSCPLL increment, independent of the colour detection bit of RTCI input.

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APPLICATION INFORMATION



- (1) Typical value.
- (2) For 100/100 colour bar.
- (3) Order no. 4312 065 02341.

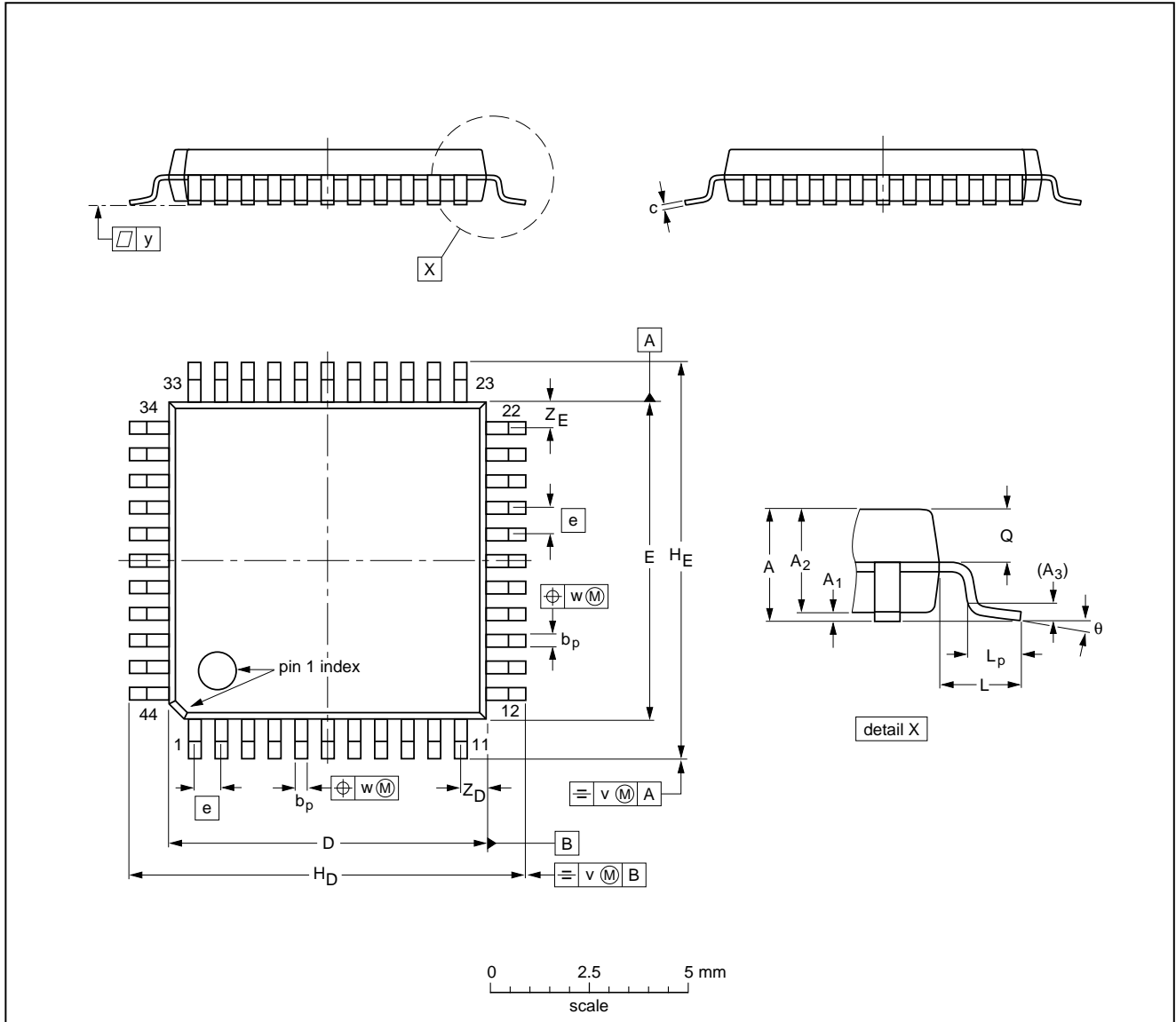
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PACKAGE OUTLINE

QFP44: plastic quad flat package; 44 leads (lead length 1.3 mm); body 10 x 10 x 1.75 mm

SOT307-2



DIMENSIONS (mm are the original dimensions)

| UNIT | A max. | A ₁ | A ₂ | A ₃ | b _p | c | D ⁽¹⁾ | E ⁽¹⁾ | e | H _D | H _E | L | L _p | Q | v | w | y | Z _D ⁽¹⁾ | Z _E ⁽¹⁾ | θ |
|------|--------|----------------|----------------|----------------|----------------|--------------|------------------|------------------|-----|----------------|----------------|-----|----------------|--------------|------|------|-----|-------------------------------|-------------------------------|-----------|
| mm | 2.10 | 0.25 0.05 | 1.85 1.65 | 0.25 | 0.40 0.20 | 0.25 0.14 | 10.1 9.9 | 10.1 9.9 | 0.8 | 12.9 12.3 | 12.9 12.3 | 1.3 | 0.95 0.55 | 0.85 0.75 | 0.15 | 0.15 | 0.1 | 1.2 0.8 | 1.2 0.8 | 10° 0° |

Note

1. Plastic or metal protrusions of 0.25 mm maximum per side are not included.

| OUTLINE VERSION | REFERENCES | | | | EUROPEAN PROJECTION | ISSUE DATE |
|-----------------|------------|-------|------|--|---------------------|----------------------|
| | IEC | JEDEC | EIAJ | | | |
| SOT307-2 | | | | | | 92-11-17 95-02-04 |

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SOLDERING

Introduction

There is no soldering method that is ideal for all IC packages. Wave soldering is often preferred when through-hole and surface mounted components are mixed on one printed-circuit board. However, wave soldering is not always suitable for surface mounted ICs, or for printed-circuits with high population densities. In these situations reflow soldering is often used.

This text gives a very brief insight to a complex technology. A more in-depth account of soldering ICs can be found in our *"IC Package Databook"* (order code 9398 652 90011).

Reflow soldering

Reflow soldering techniques are suitable for all QFP packages.

The choice of heating method may be influenced by larger plastic QFP packages (44 leads, or more). If infrared or vapour phase heating is used and the large packages are not absolutely dry (less than 0.1% moisture content by weight), vaporization of the small amount of moisture in them can cause cracking of the plastic body. For more information, refer to the Drypack chapter in our *"Quality Reference Handbook"* (order code 9397 750 00192).

Reflow soldering requires solder paste (a suspension of fine solder particles, flux and binding agent) to be applied to the printed-circuit board by screen printing, stencilling or pressure-syringe dispensing before package placement.

Several techniques exist for reflowing; for example, thermal conduction by heated belt. Dwell times vary between 50 and 300 seconds depending on heating method. Typical reflow temperatures range from 215 to 250 °C.

Preheating is necessary to dry the paste and evaporate the binding agent. Preheating duration: 45 minutes at 45 °C.

Wave soldering

Wave soldering is **not** recommended for QFP packages. This is because of the likelihood of solder bridging due to closely-spaced leads and the possibility of incomplete solder penetration in multi-lead devices.

If wave soldering cannot be avoided, the following conditions must be observed:

- **A double-wave (a turbulent wave with high upward pressure followed by a smooth laminar wave) soldering technique should be used.**
- **The footprint must be at an angle of 45° to the board direction and must incorporate solder thieves downstream and at the side corners.**

Even with these conditions, do not consider wave soldering the following packages: QFP52 (SOT379-1), QFP100 (SOT317-1), QFP100 (SOT317-2), QFP100 (SOT382-1) or QFP160 (SOT322-1).

During placement and before soldering, the package must be fixed with a droplet of adhesive. The adhesive can be applied by screen printing, pin transfer or syringe dispensing. The package can be soldered after the adhesive is cured.

Maximum permissible solder temperature is 260 °C, and maximum duration of package immersion in solder is 10 seconds, if cooled to less than 150 °C within 6 seconds. Typical dwell time is 4 seconds at 250 °C.

A mildly-activated flux will eliminate the need for removal of corrosive residues in most applications.

Repairing soldered joints

Fix the component by first soldering two diagonally-opposite end leads. Use only a low voltage soldering iron (less than 24 V) applied to the flat part of the lead. Contact time must be limited to 10 seconds at up to 300 °C. When using a dedicated tool, all other leads can be soldered in one operation within 2 to 5 seconds between 270 and 320 °C.

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DEFINITIONS

| | |
|---|---|
| Data sheet status | |
| Objective specification | This data sheet contains target or goal specifications for product development. |
| Preliminary specification | This data sheet contains preliminary data; supplementary data may be published later. |
| Product specification | This data sheet contains final product specifications. |
| Limiting values | |
| Limiting values given are in accordance with the Absolute Maximum Rating System (IEC 134). Stress above one or more of the limiting values may cause permanent damage to the device. These are stress ratings only and operation of the device at these or at any other conditions above those given in the Characteristics sections of the specification is not implied. Exposure to limiting values for extended periods may affect device reliability. | |
| Application information | |
| Where application information is given, it is advisory and does not form part of the specification. | |

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Purchase of Philips I²C components conveys a license under the Philips' I²C patent to use the components in the I²C system provided the system conforms to the I²C specification defined by Philips. This specification can be ordered using the code 9398 393 40011.

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